

- PSoc & Switched Cap

if PSoc etc. piece doesn't work - the other piece should
modular

& idea w/ tech.

MPPT

mpja

digikay

anemom

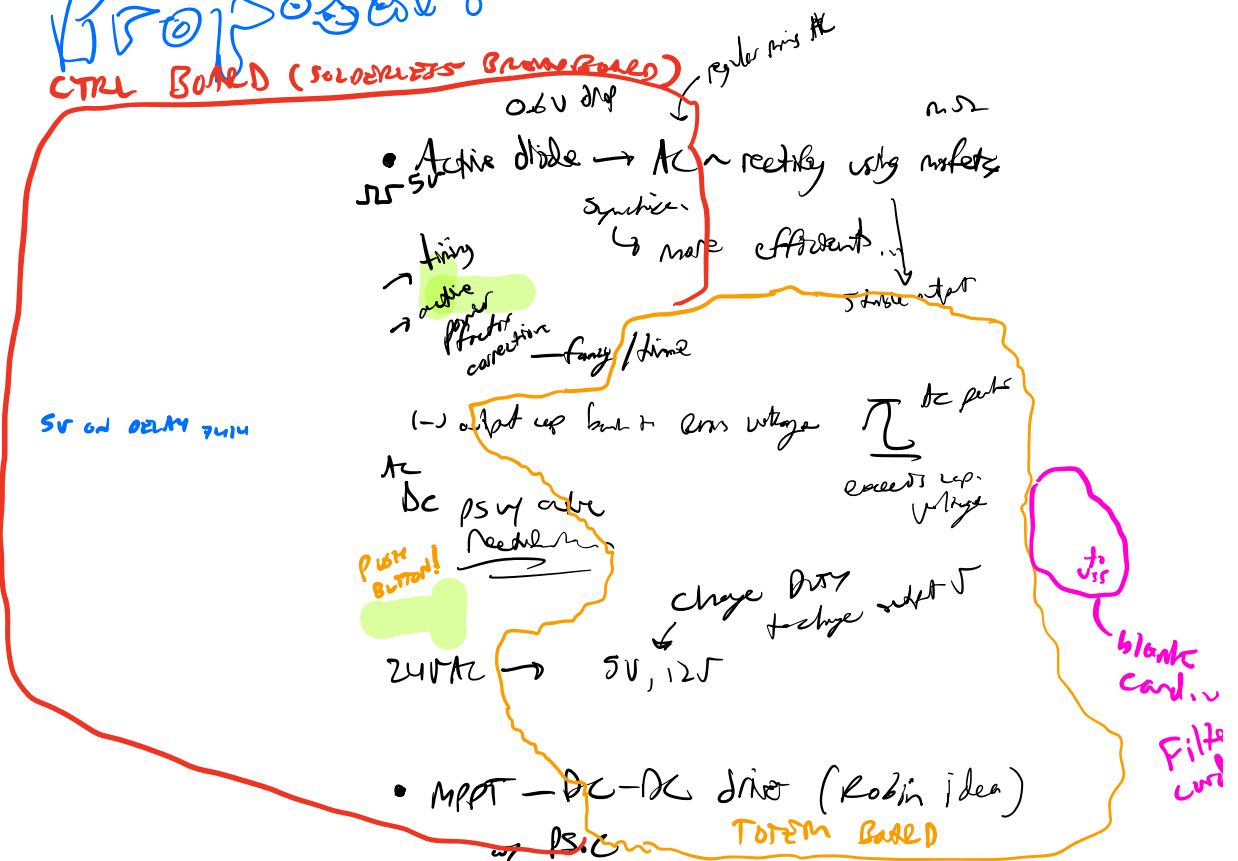
allelectronics

sparkfun

adafruit

Project Proposal!

CTRL BOARD (SOLAR PANE BASED)



Design
Kerns 1:

Pool pump → largest stack?

2) sufficient pump ← largest stack

Performance?

* Precision *

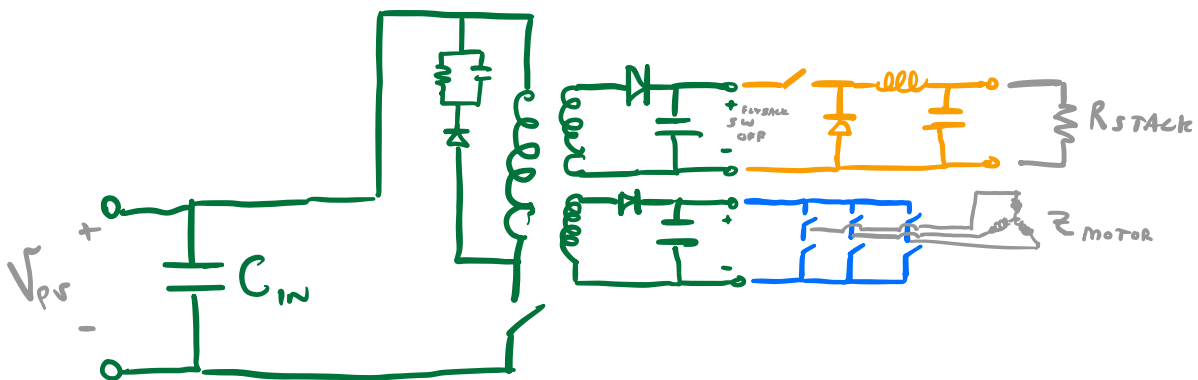
CAPEX

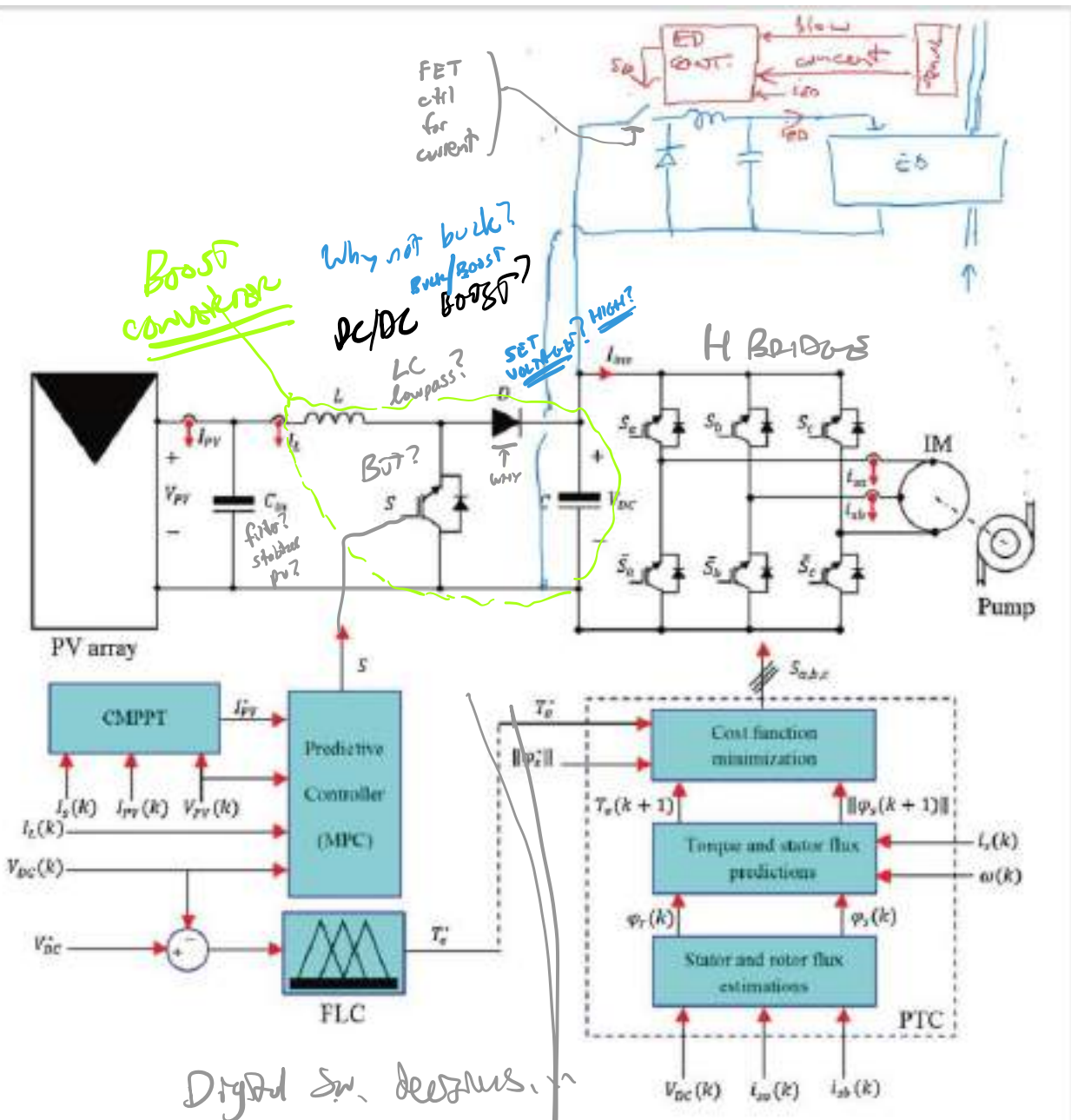
BANDWIDTH

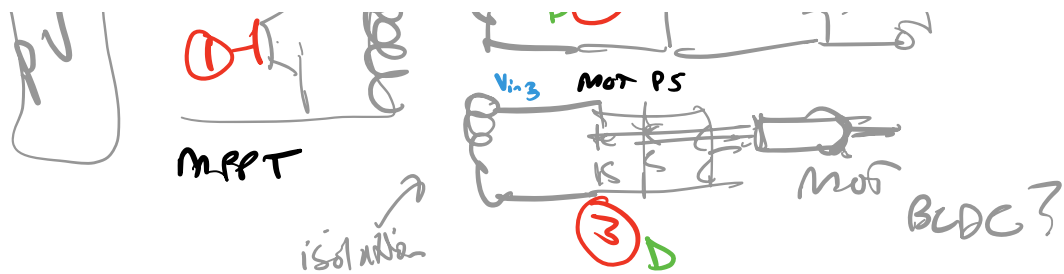
SEC

①

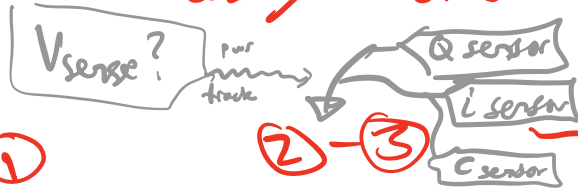
②







ctrl gets 1, 2, 3 from digital signal to analog to driver.



✓ Power sw for smaller 2, C

① from power today?

MPPT alg.?

how does this work?

CCM

②-③ from I/s/m / model based?

2, 3 as ctrl for MPPT adjust

PSOL for digital control (speeds? freq.?)

What happens if I sw body?

@ 2, 3?

→ V just pulled down from whatever is there ~?

→ can't ever pull it as V_{in} ~?

"full D.R."

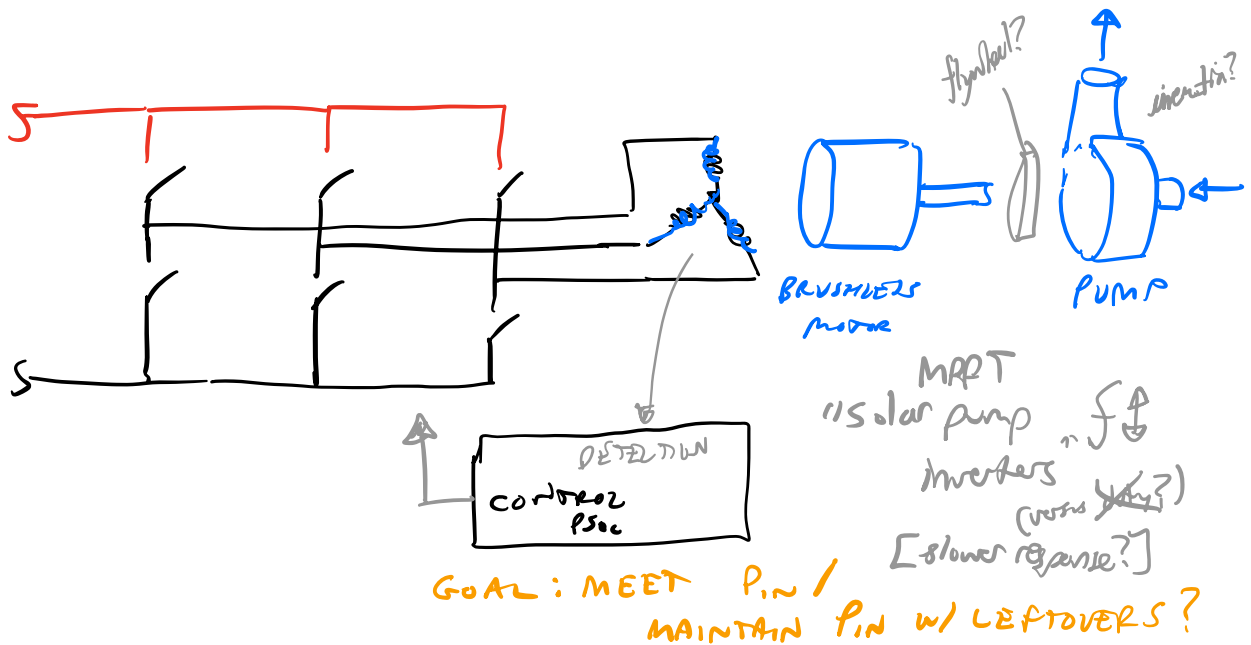
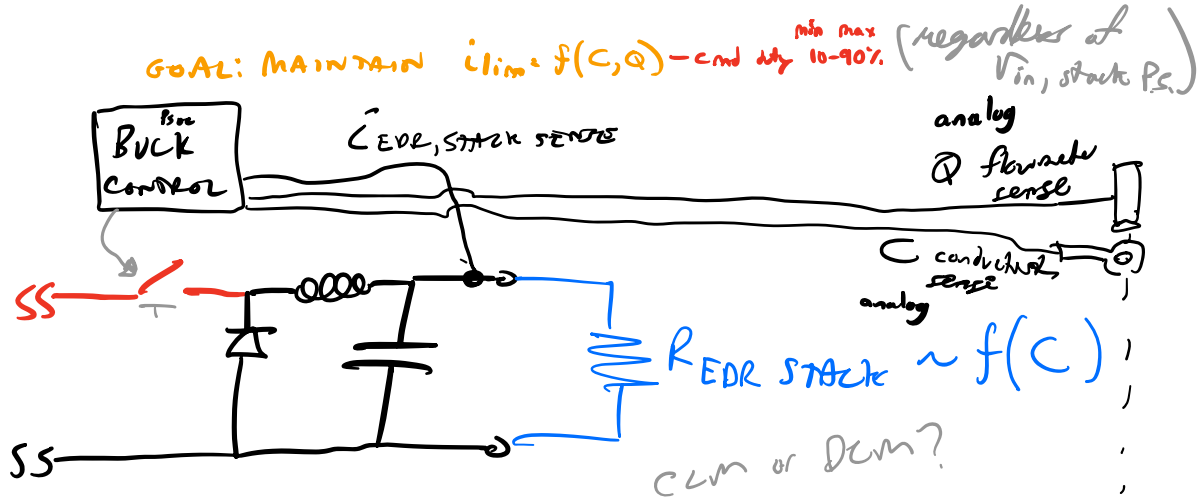
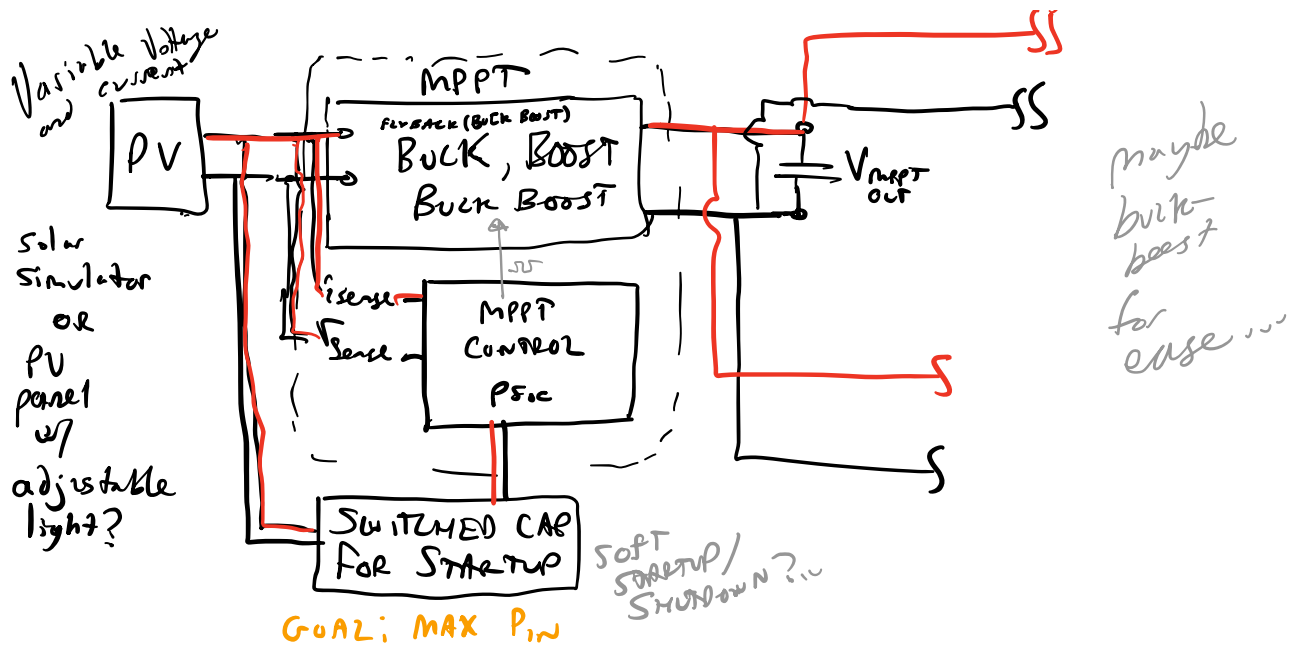
How do we decide 1, 2, 3 speeds? ~

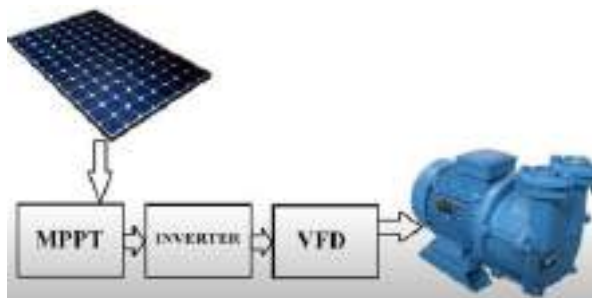
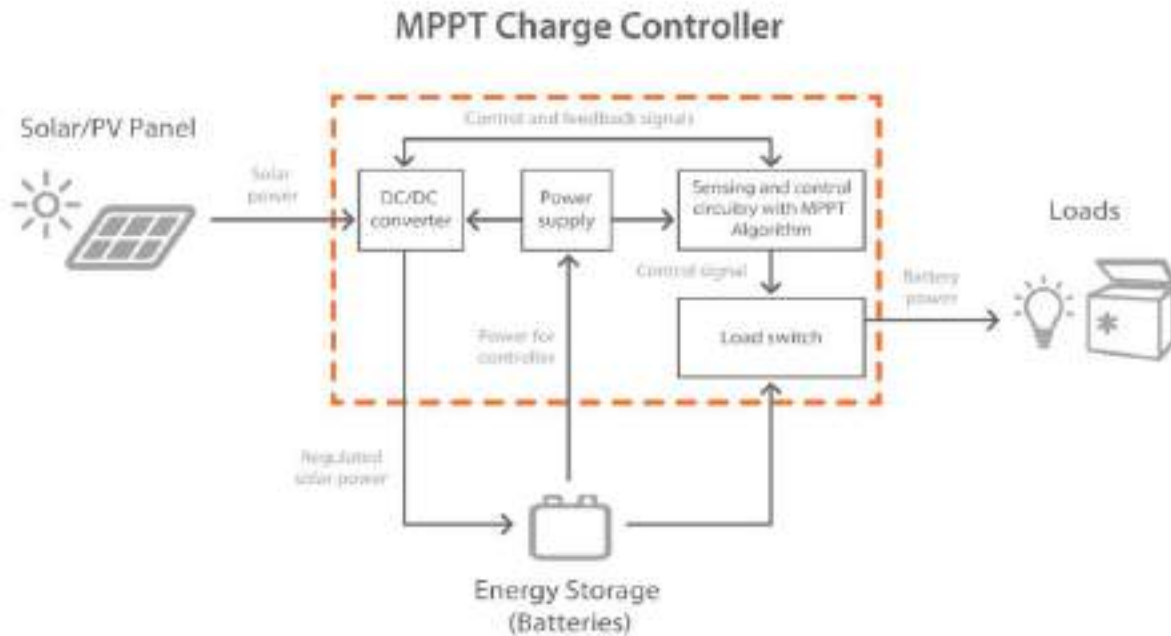
{ Do @ small scale to prove can work ~
→ scaling args for large scale }

REQ / RATION

(+) coupled ctrl

Aim is TO PROVE CTRL → NOT TO OPT, ~





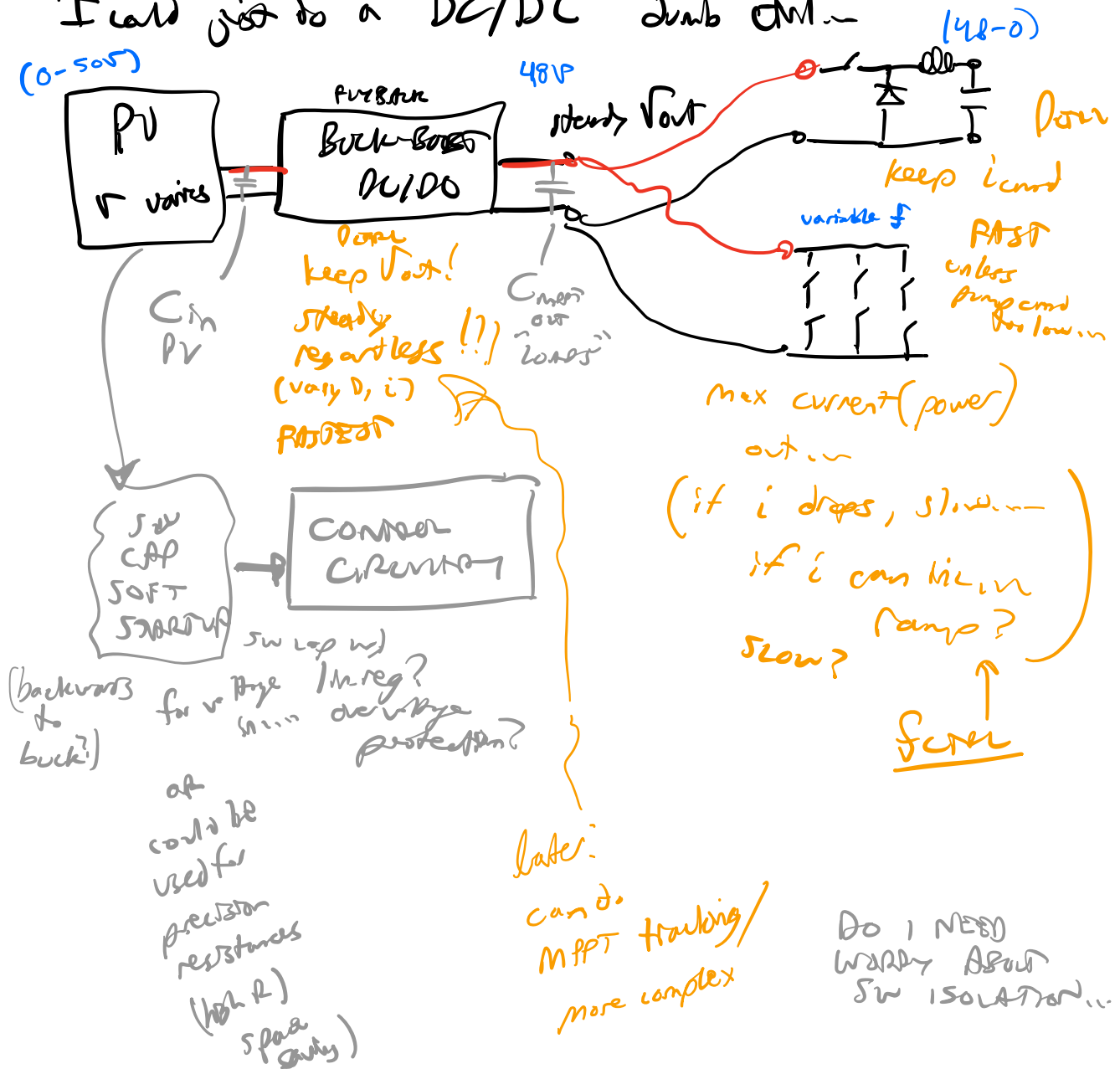
IGBT > MOSFET
 for π , high current
 high V
 get fast switching
 heat dissipation !!!



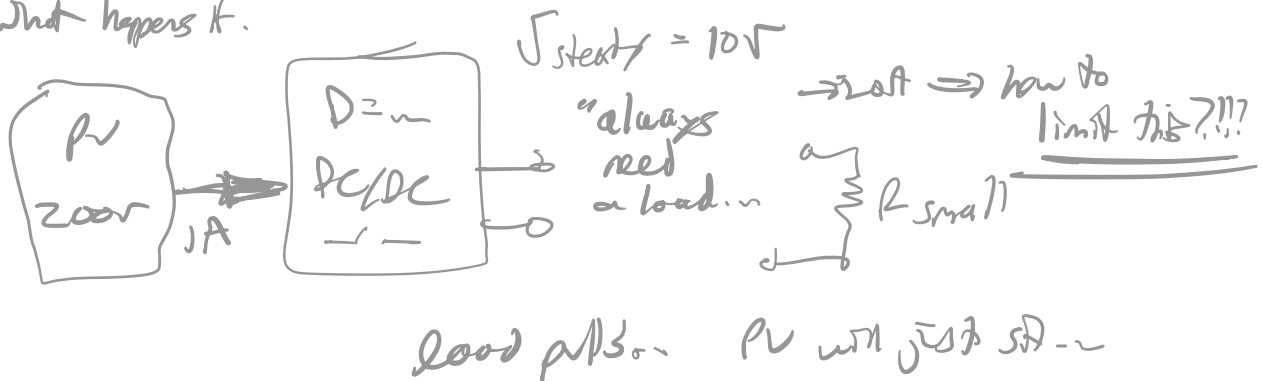
https://www.youtube.com/watch?v=NLrpxi5LqYw&ab_channel=MediElectronics

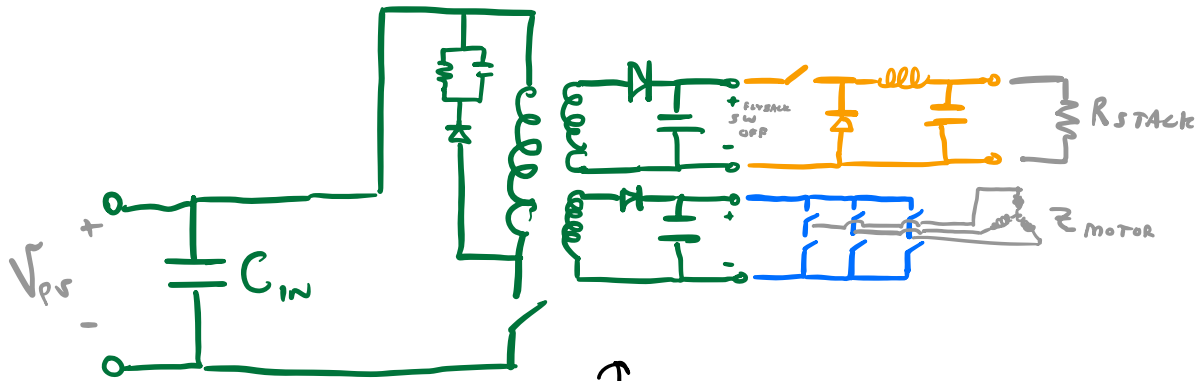
□ input voltage regulation (max power point)
 □ output V regulation $\leftarrow$ feedback circuit...

I could just do a DC/DC dumb ctrl.



What happens if.

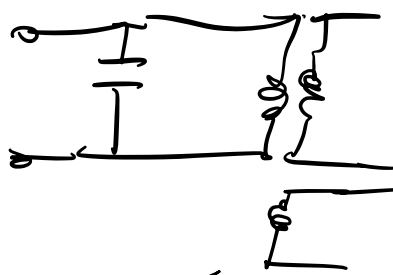




↑
related
flyback anem? 1:1

↑
pg all
stack?

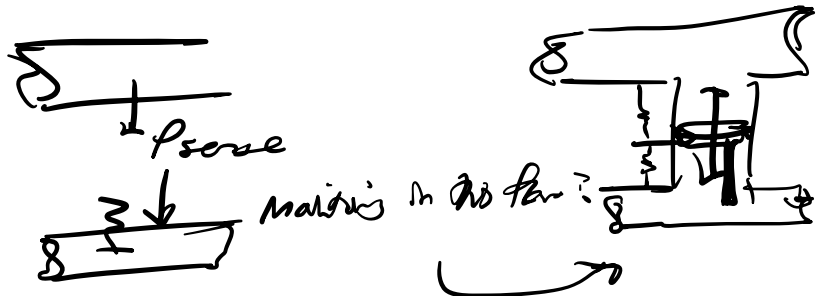
↑
just one HV rail?



only 2 for lines
T, D
carbon
pump w/
extraction
simple?
... elm chl
Rotor
↓ matched?
pump
chl fast? ...

24VDC → valve, chl, e-rod, etc

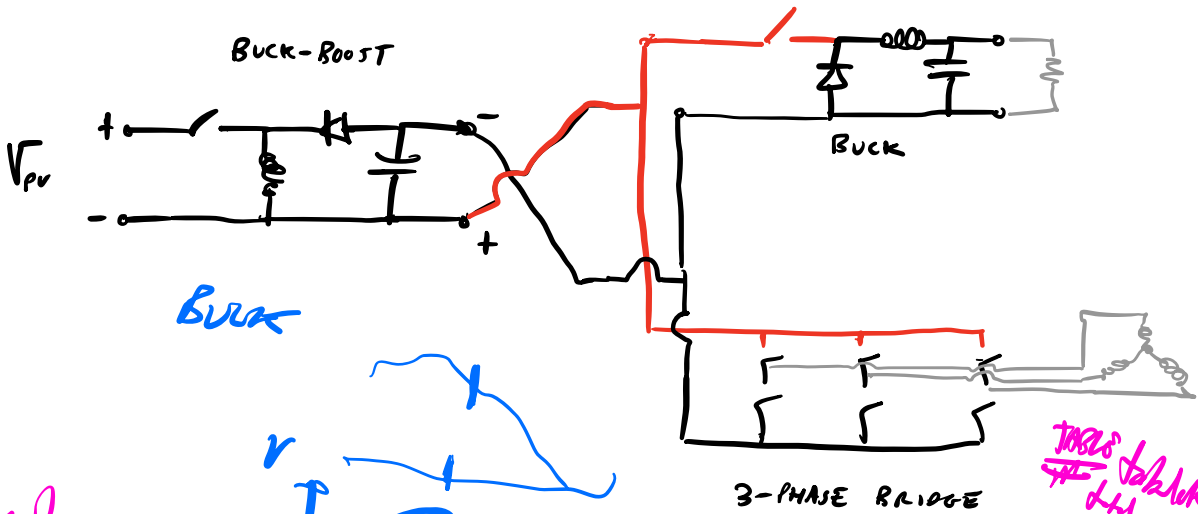
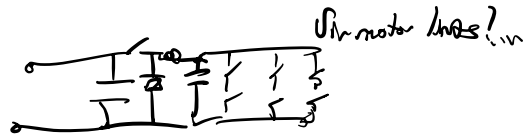
↑
pasive
self-regulating
flow?



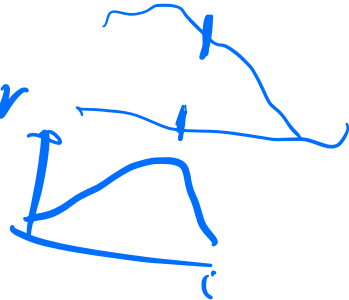
main in the fan?

Pro.

Prog. value sets flow based on other stream...



w/ panel info →



4x12V

50V

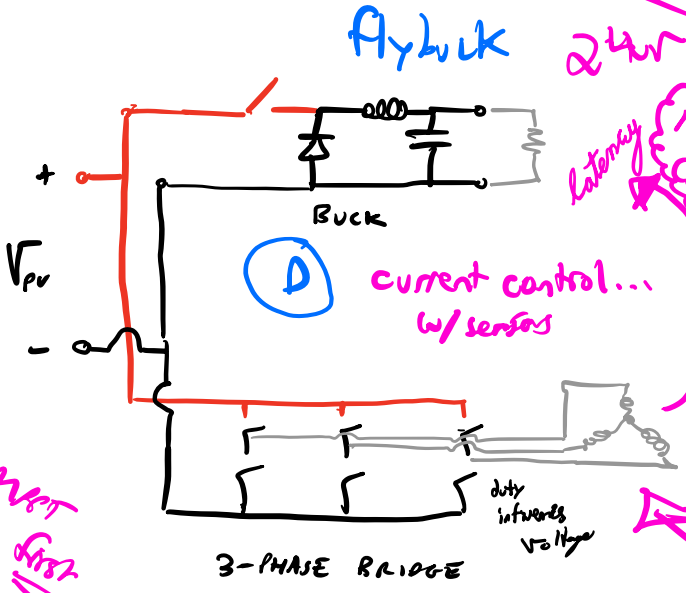
V_{opt}

~ 80V Voc

Met first

"two load in MPPT"
"multi load MPPT"

Differs than 1A



3-PHASE BRIDGE

ⓓ

current control... w/ sensors

latency
Master
Control
Loop

f sw

ⓓ

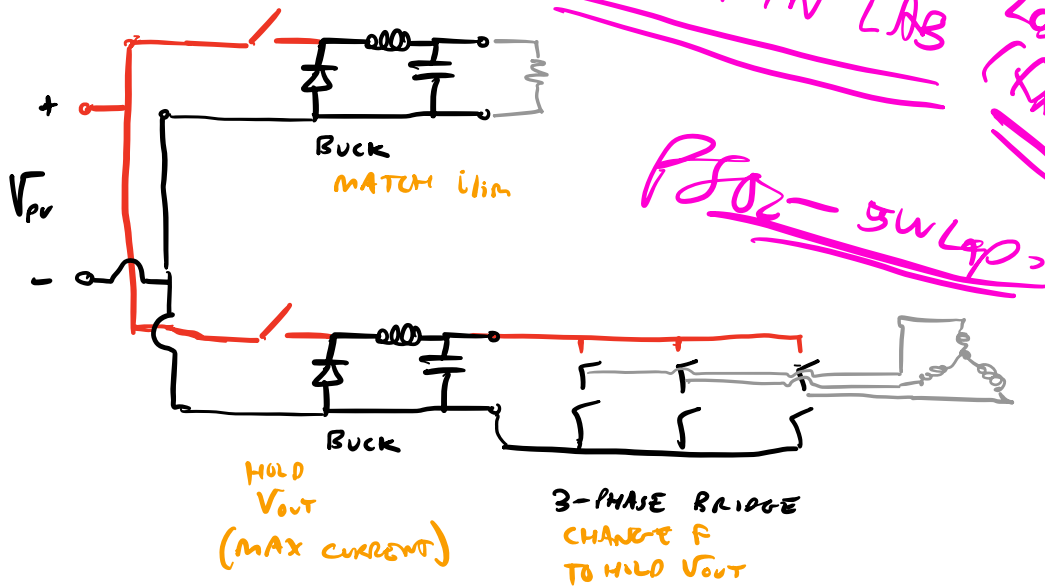
hold bus voltage...

of pump...

CONVERTERS

NO LIQ IN LAB

ANALOG
HEAVY
LOAD
(Fm) etc



Get ratings / DRS (AV etc.)

V_{in}

Model Load!

$V_{o \text{ stack}}$ 0-36VDC $\rightarrow$ 1A?
 $V_{o \text{ pump}}$ 24VDC? 2A

Analog programmable
variable resistor?

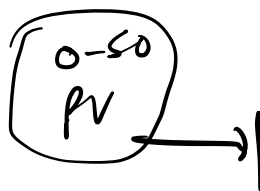
static behavior is

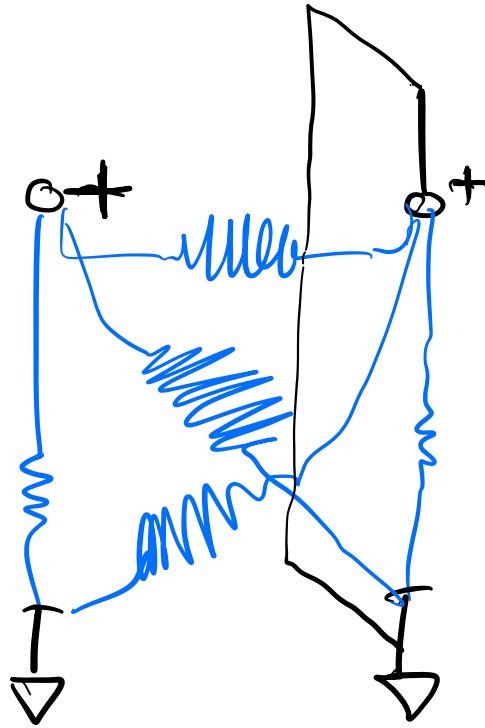
C, L, Q

AC R
Circuit

for home side
just 2 bricks

Psic power? -



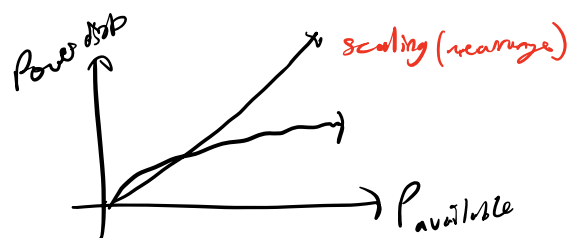
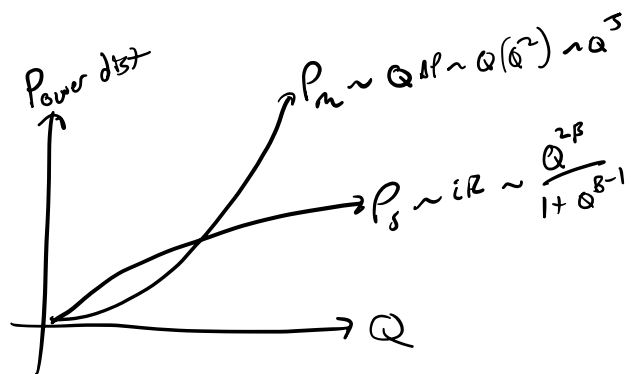
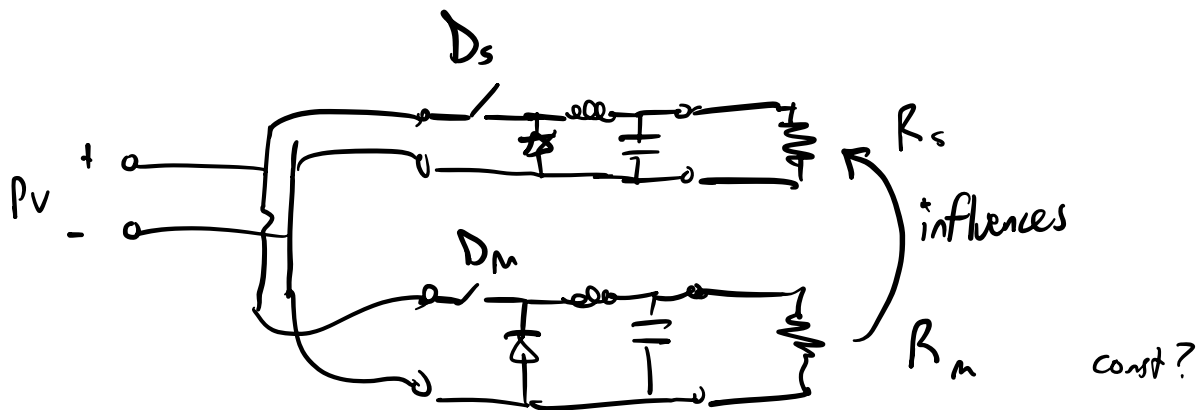
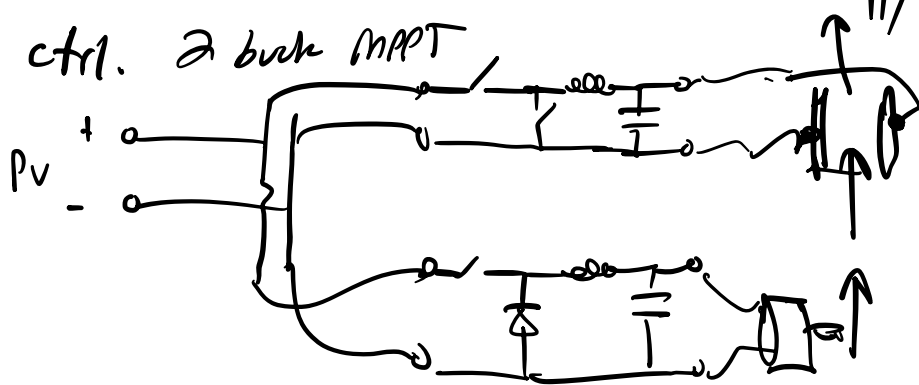


[https://www.ti.com/lit/an/snva674c/snva674c.pdf?](https://www.ti.com/lit/an/snva674c/snva674c.pdf?ts=1698876769581&ref_url=https%253A%252F%252Fwww.bing.com%252F)

[ts=1698876769581&ref_url=https%253A%252F%252Fwww.bing.com%252F](https://www.ti.com/lit/an/snva674c/snva674c.pdf?ts=1698876769581&ref_url=https%253A%252F%252Fwww.bing.com%252F)

Fly Buck Converter

ideal ctrl. 2 buck MPPT 11/14/23

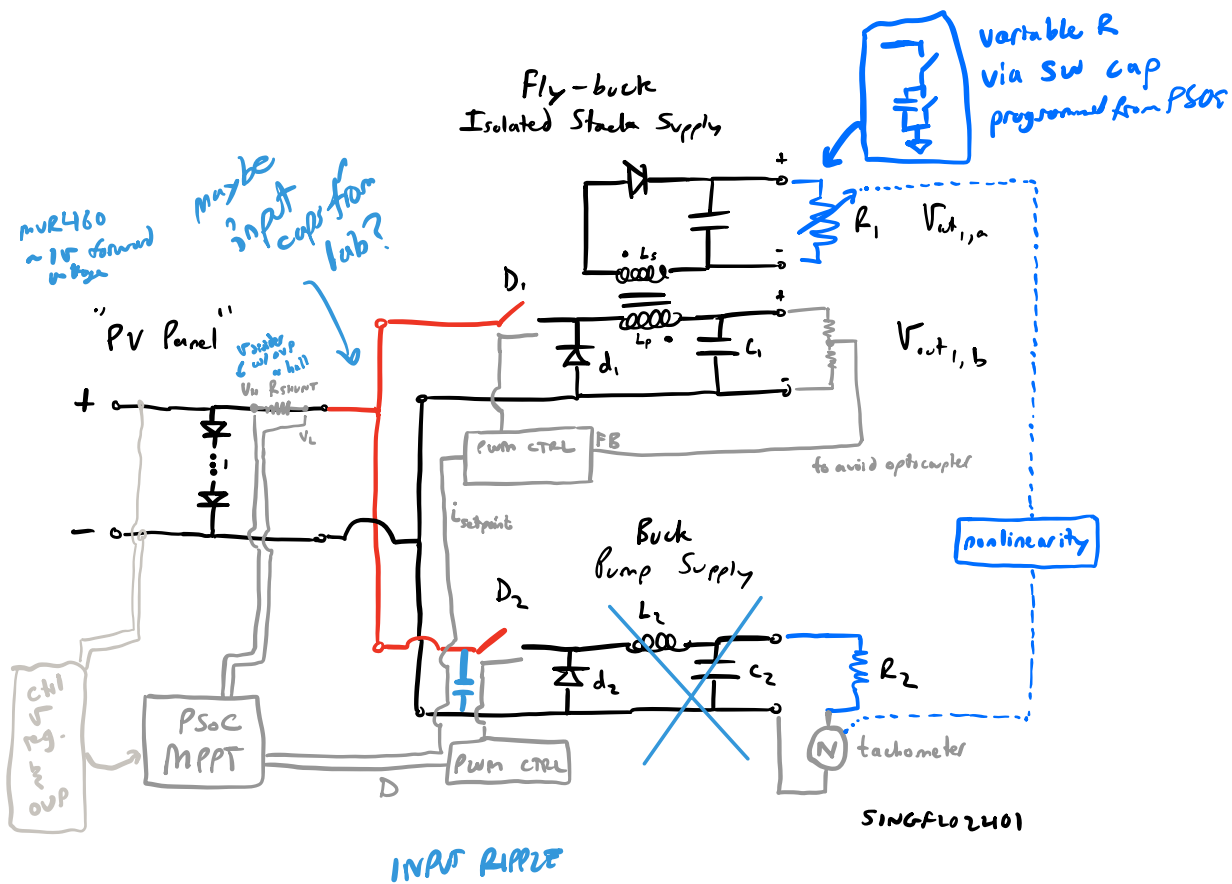
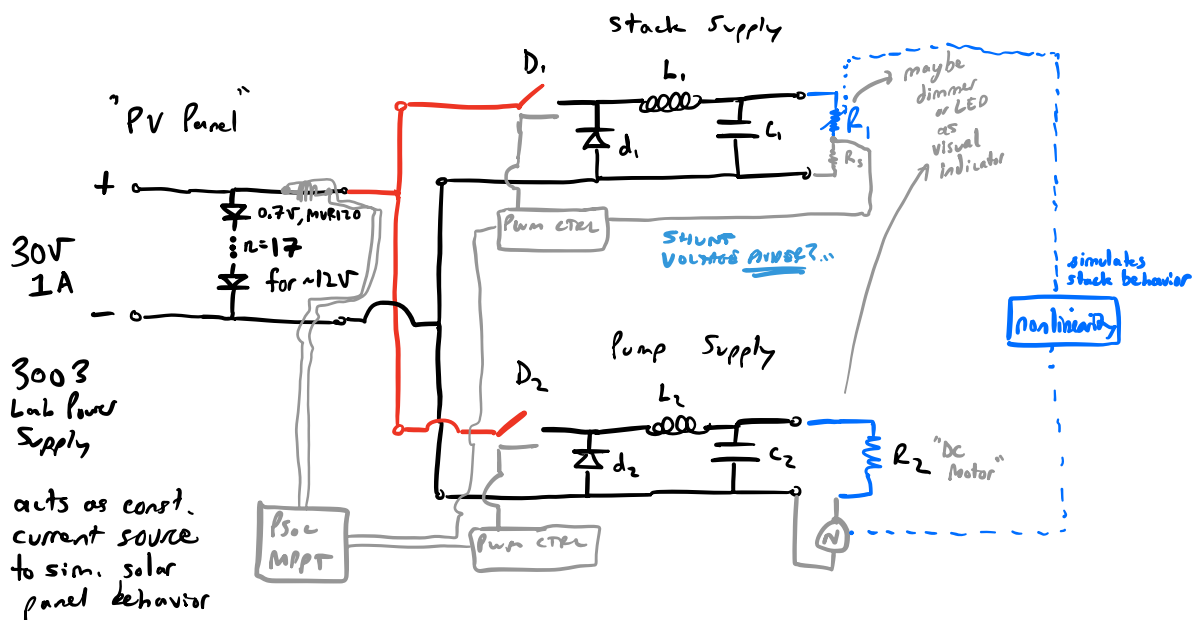


$P = i^2 R$

to model expected R variance (optimal power distribution)

Power Available

ARCHITECTURE



DESIGN REQUIREMENTS

small scale EDR

• 24V DC (nominally) pump w/ $\sim 12V-24V$ range

• max current = **0.9A** $\sim 26\Omega$

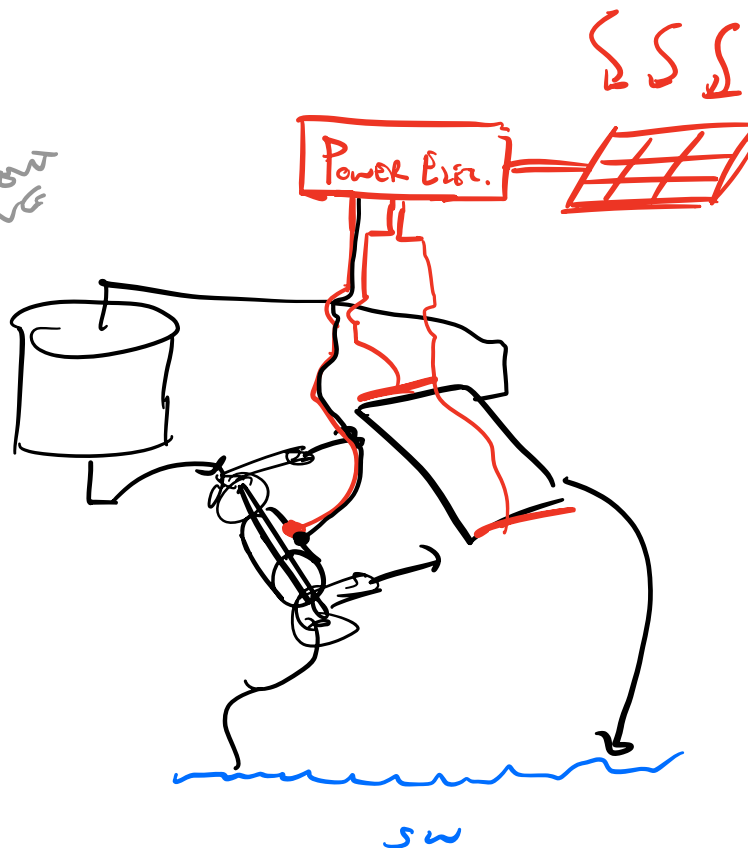
• $\Delta V = \cancel{0.5V}$ **100mV**
ripple

• variable voltage & current stacks

• Δi ripple = **0.1A** (based on factor of safety, i_{lin})
 $\Delta V = 100mV$
 $\sim 30\Omega$

from Melissa $\rightarrow$ 25 for this supply...
 $\sim 0-30V$ range
 $\sim 0-2A$ range
250mA
 $\downarrow$
ideally 10A for servomotor

$\downarrow$
L, C,
FET CURRENT
LIMITING



	0-60 V_{max} μ 25-24	V_{max} X	Cont I_{max} 0.5-0.6	I_{max}
SMRKS				
PMP	0-60 V_{max} μ 0-24	V_{max} 100 mt	X	I_{max}
CTRL	0-60 V_{max} μ 11N or 5, 12	V_{max} X	I_{max} X	I_{max}

related to GORDC
SA

BUCK SMUG P.S. CONVERTER SIZING

FET ✓ current good any in class work
 diode ✓ current good 460

lets choose L ... then C ...

CCM: $\frac{\Delta i}{2} \leq \langle i_L \rangle$ $V_L = L \frac{\Delta i}{\Delta t}$

$\langle i_L \rangle = \langle i_o \rangle$
 buck

$$L \geq \frac{V_L \Delta t}{2 \langle i_o \rangle} = \frac{V_o (1-D) T}{2 \langle i_o \rangle}$$

D, $\langle i_o \rangle$

during worst times ...

T choose $\frac{1}{30kHz}$
for familiarity ...

100mA ~ L largest val.
 $\langle i_o \rangle$ smallest ←
 D smallest ... = 0

V_o largest = 30V

$L = 5mH$
 $T = \frac{1}{30000}$

$L = 1.5mH$
 $T = \frac{1}{100,000}$

from i_{lin}
 @ lowest
 conc.
 @ lin velocity
 { ~ 200 ppm }
 { ~ 2 cm/s }

$\Delta V = \frac{1}{8} \Delta i T \frac{1}{C}$

from $q = \frac{1}{2} \langle i_{out} \rangle \frac{T}{2}$
 $\Delta i/2$



$$C \geq \frac{1}{8} \Delta i \frac{T}{\Delta V}$$

$$\Delta i \approx 2 \langle i_2 \rangle \approx 2 \langle i_o \rangle$$

$$\approx 2 (0.1 \text{ A})$$

min
i_{output}

$$C \geq \frac{1}{8} (0.2 \text{ A}) \frac{\left(\frac{1}{100,000} \right)}{0.1 \text{ V}}$$

← min f

← min ripple

w/ max current: 3A, 5A

$$\Delta i \approx 0.2 \text{ A}$$

$$\Delta V_{\text{spec}} \sim 100 \text{ mV} \sim 0.1 \text{ V}$$

$$\underline{C \geq 2.5 \mu\text{F}} \quad @ \text{ up to } 60 \text{ V to be safe.}$$

if $T = 30,000$ @ $\Delta i \approx 200 \text{ mA}$ and $\Delta V \sim 50 \text{ mV}$

$$\underline{C \geq 17 \mu\text{F}} \quad \text{capacitor not an issue.}$$

if $C \sim 1000 \mu\text{F}$ then

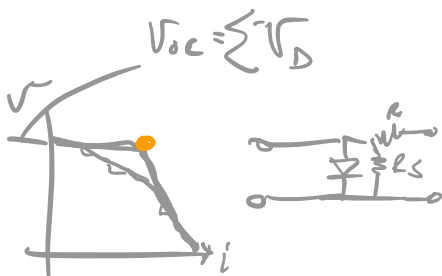
220 μF → 0.011 V
11 mV

$$\Delta V \sim \frac{1}{8} (0.2 \text{ A}) \left(\frac{\frac{1}{10,000}}{1 \text{ mF}} \right)$$

$$\sim 0.0025 \text{ V} \quad \sim 2 \text{ mV}$$

w/ 5A 10 kHz 220 μF → 0.284 V ripple

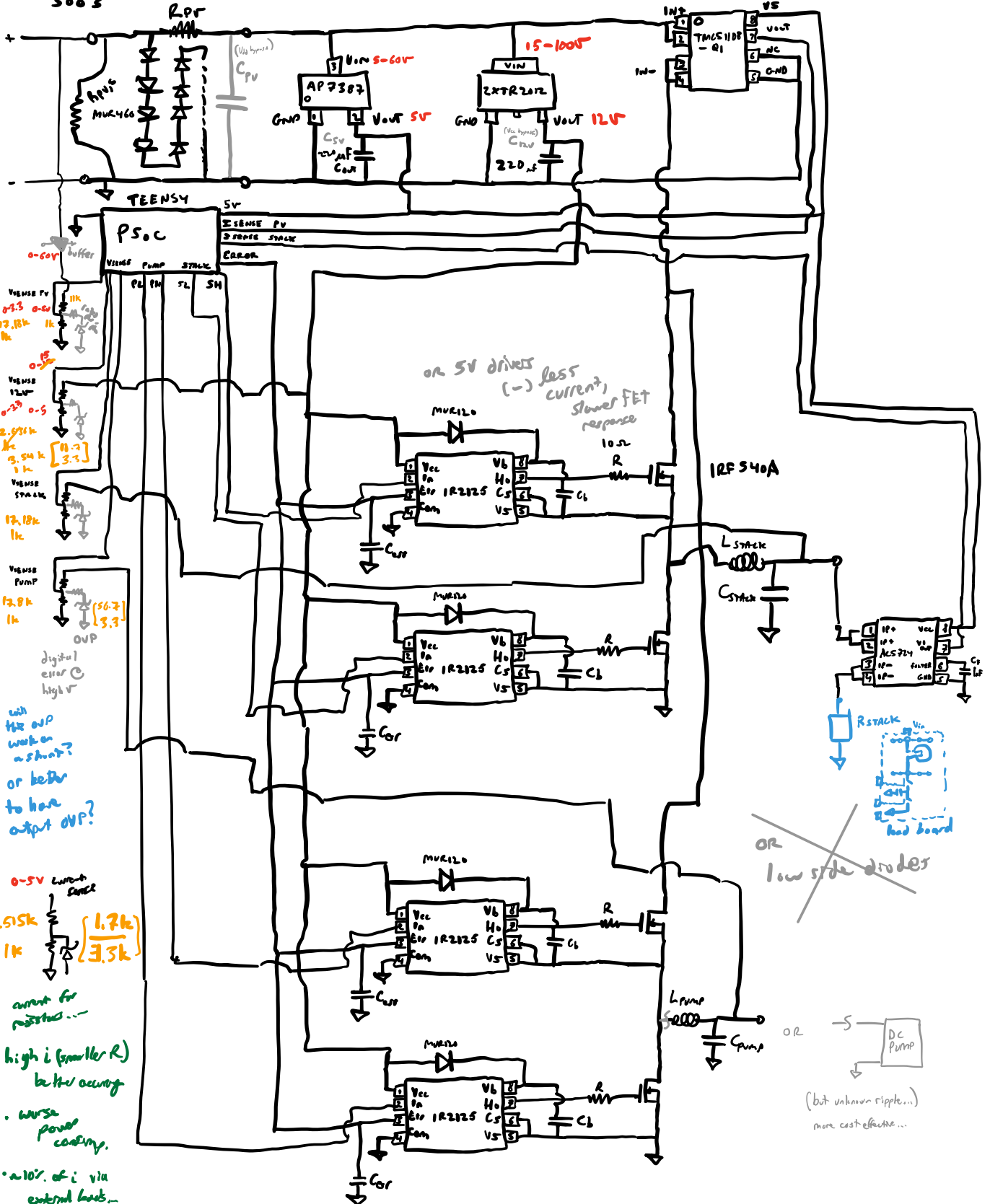
SA 10 kHz 1000 μF → 0.0625 V ripple





$\frac{1}{8}$ to 1 watt most need
need power
resistor

PS
3003



OR 5V drives
(-) less
current,
slower FET
response

OR
low side driver

OR
DC PUMP

(but unknown ripple...)
more cost effective...

with
the OVP
works on
a short?
or better
to have
output OVP?

0-5V current
sense
 $1k$ $\left[\frac{1.7k}{3.3k} \right]$

current for
resistor...

• high i (smaller R)
better accuracy

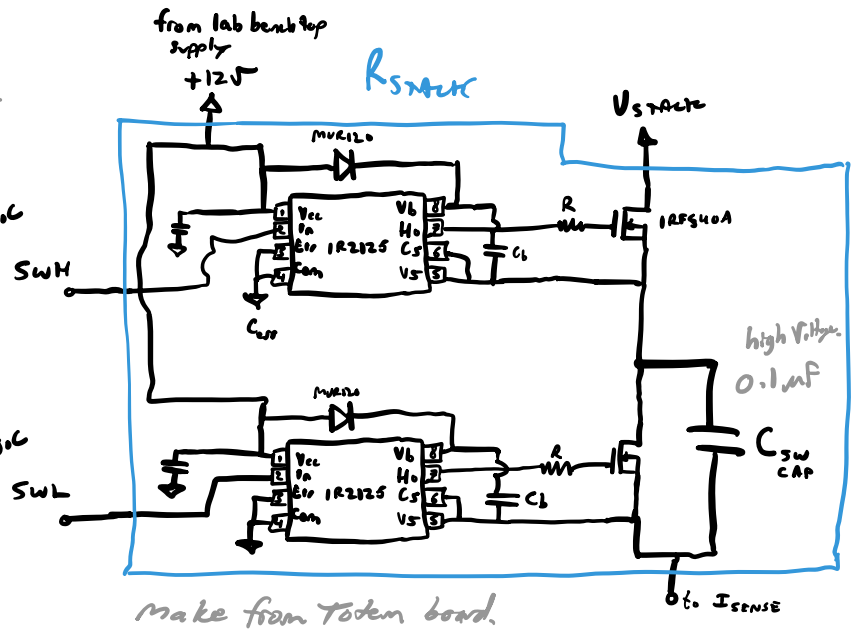
• worse
power
constr.

• a 10% of i via
external load...

just PSoc for write R? - comfortably...
 0.1 μ F $\rightarrow$ [0-10 kHz]

(1k to 100k) $f_{10,000}$ f_{100}

f_{ctrl} from PSoc
 SWH
 SWL
 screen demands



• PSoc ground unused DO pins (or assign them)

- digital output limited to ~ 0 or $5V$ 10kHz

- expected input PSoc digital $\square$ 0 to

analog $\sim$ 0 to ~~5V~~ 4V
 experimentally reliable (comparator)

could you drive a μ FT directly w/ PSoc? load?

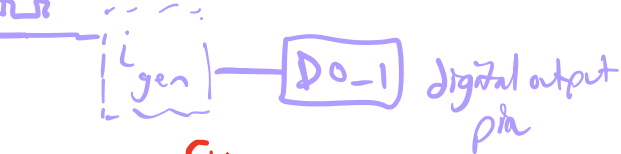
from digital out pins?

waveform



V_{ref}
 (duty)

add max. function...



current is limited... $\rightarrow$ BJT



• ways to visualize on PSoC?
 (scope) *seems like impossible.*

• explain drive modes
 (strong, high impedance)
 ✓ *etc*

• analog output? *where?* **ADC**

• V sense & OVP

• general circuit review

Shunt Resistance (R_{sh}) and Series Resistance (R_s)

During operation, the efficiency of PV cells is reduced by the dissipation of power across internal resistances. These parasitic resistances can be demonstrated as a parallel shunt resistance (R_{sh}) and series resistance (R_s) as depicted below.

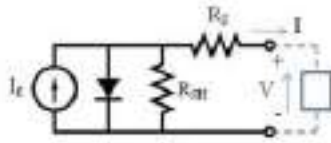


Figure 8 - Equivalent Circuit of a PV Cell

Decreasing R_{sh} and increasing R_s will decrease the fill factor (FF) and P_{max} as shown in Figure 9 below. If R_{sh} is decreased too much, V_{oc} will drop, while increasing R_s excessively can cause I_{sc} to drop instead.

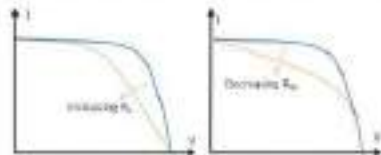


Figure 9 - Effect of Changing R_s and R_{sh} from Ideality

• more analog inputs?

{ PSoC GPIO
 is analog, digital I/O
 bidirectional.

{ try DO7
 DI1
 DI2
 AI1
 AI2 w/ an
 analog
 input... }

"DO" *(sampled to smooth...)*
 "DI" artifact of CB? *check*

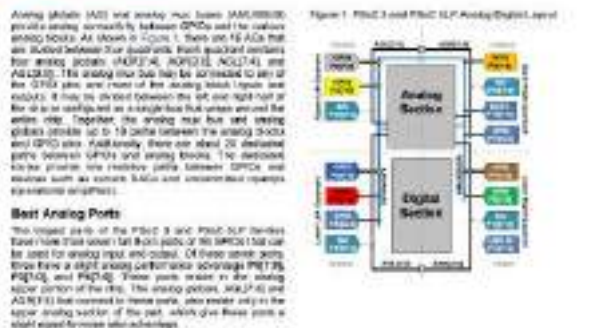
limiting

DO7 } *optocoupled*
 DI1 } *(likely not bidirectional)*

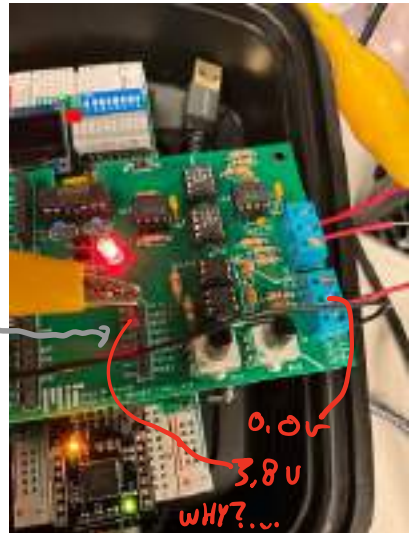
PO-4 } *happens*
 P3.2 } *ups wiring.*



Figure 1 PSoC 3 and PSoC 4LP Analog/Digital Layout



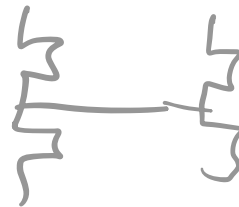
- Digital Input pins pulsed high 3.7, 3.8 V
DI1, DI2 probed



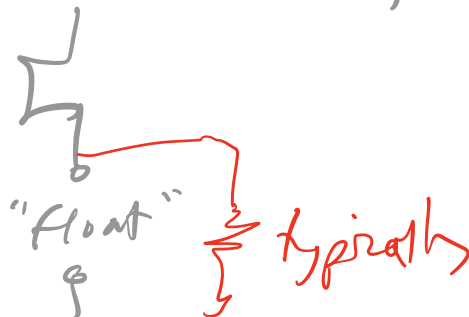
$\left[\begin{array}{l} PS - \text{pin signal} \\ PR - \text{into driver } N, P \text{ FET} \end{array} \right\} \text{ PIN to } \downarrow \text{ high or GND}$

$\left[\text{resistor pullup} : \rightarrow \text{assert low, let's it go high} \right]$

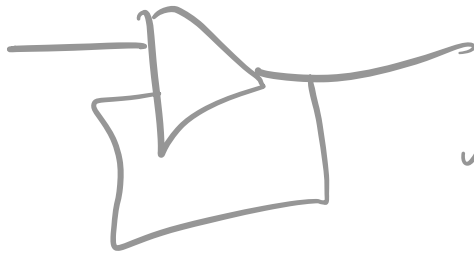
pin low



open source



high voltage



low output

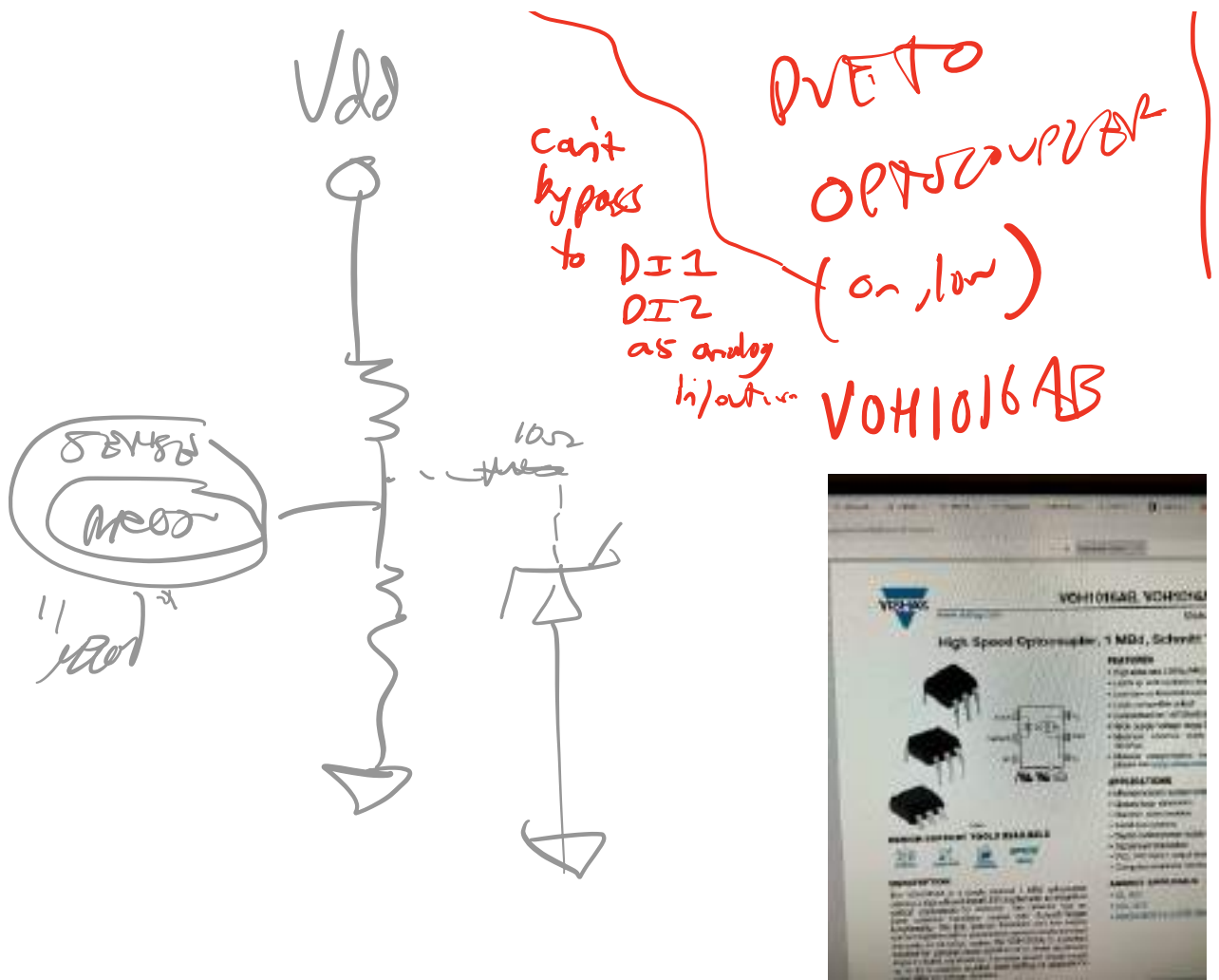
Buffer → UPDATING CURRENT

high input

WHEN TO USE:

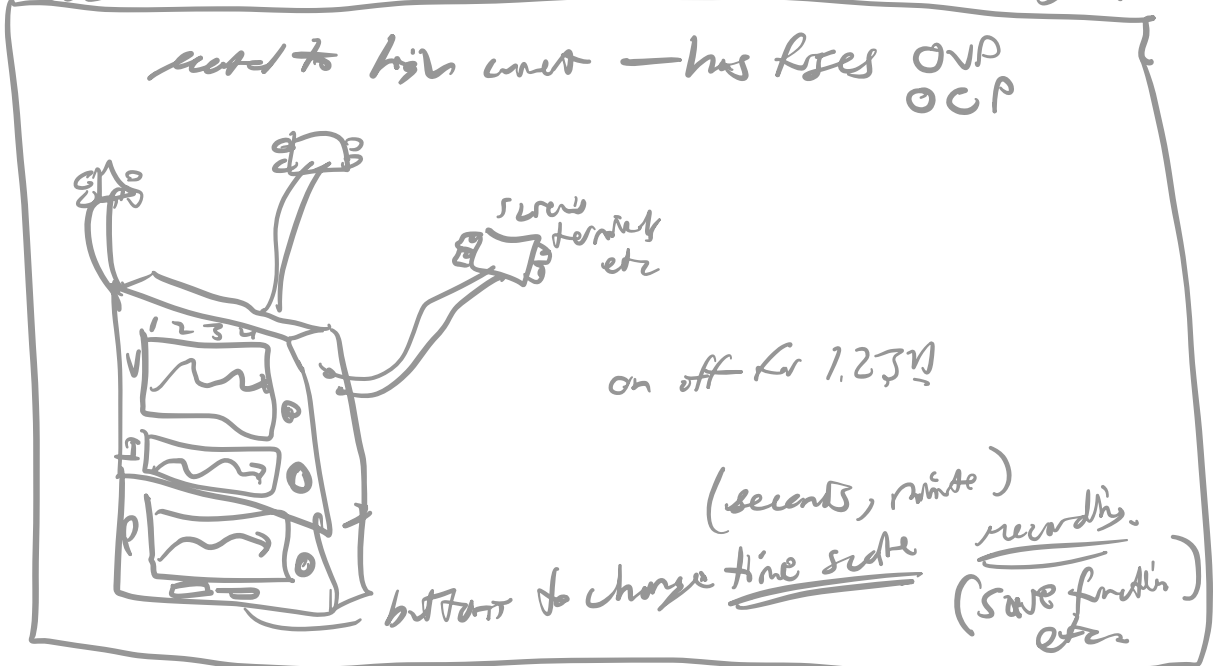
FPGA	myrio	PSOC	Microcontroller
fast CPU ↓ Analog Microcontroller	drag blocks labview	mostly digital embedded SPI interface less ↓ drag in few OP AMPs (visual blocks)	room for supporting hardware ESP32, Teensy Arduino Espressif with bluetooth product Fast to H2 (-jordan) Espressif RTOS 555 Teensy Faster analog R/D

DIGITAL INPUTS ARE INVERTED

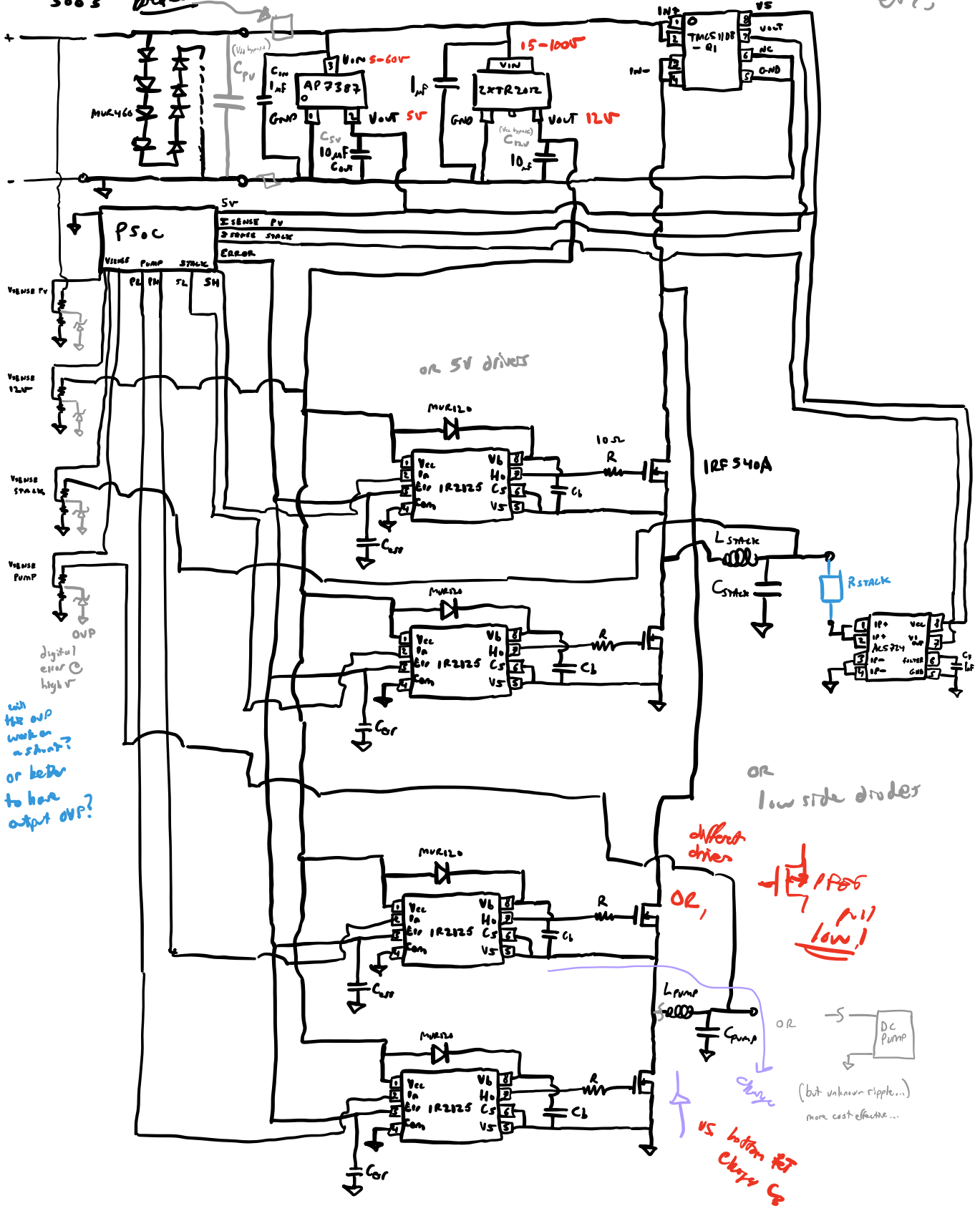


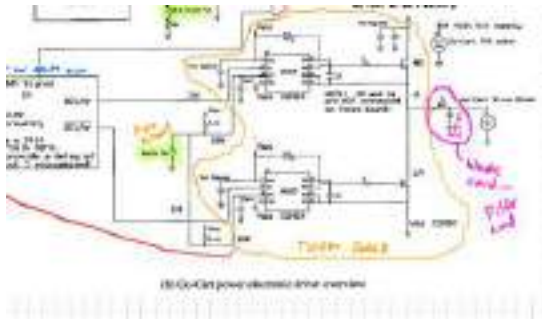
IOE "VIP" box $\Rightarrow$ For transient response checking!!

used to high unit — has rises OVP
OCP

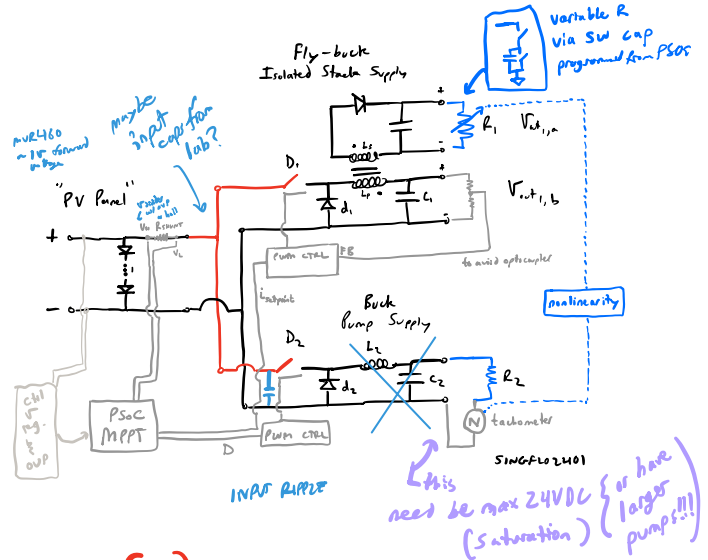


- add Aux pins / extra pins on PCB for extra daps, etc.





high side N
 V_Z
 low side P Mosfet



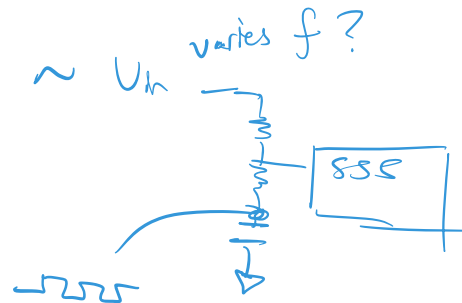
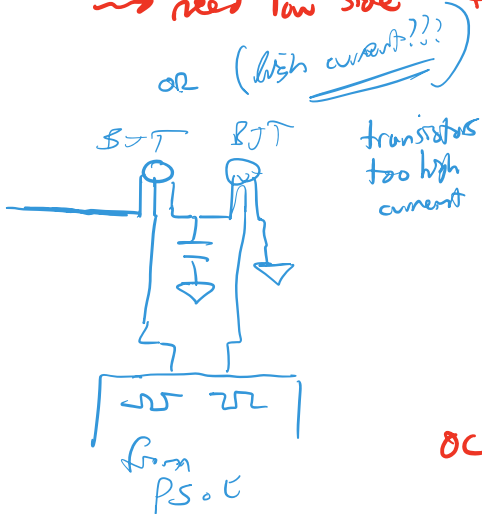
synchronous (low side FET) — (+) ΔV low side mosfet
 asynchronous (diode) less than across diode.

less power diss.
 higher efficiency

- reverse voltage
- forward voltage drop
- forward current

→ power dissipation

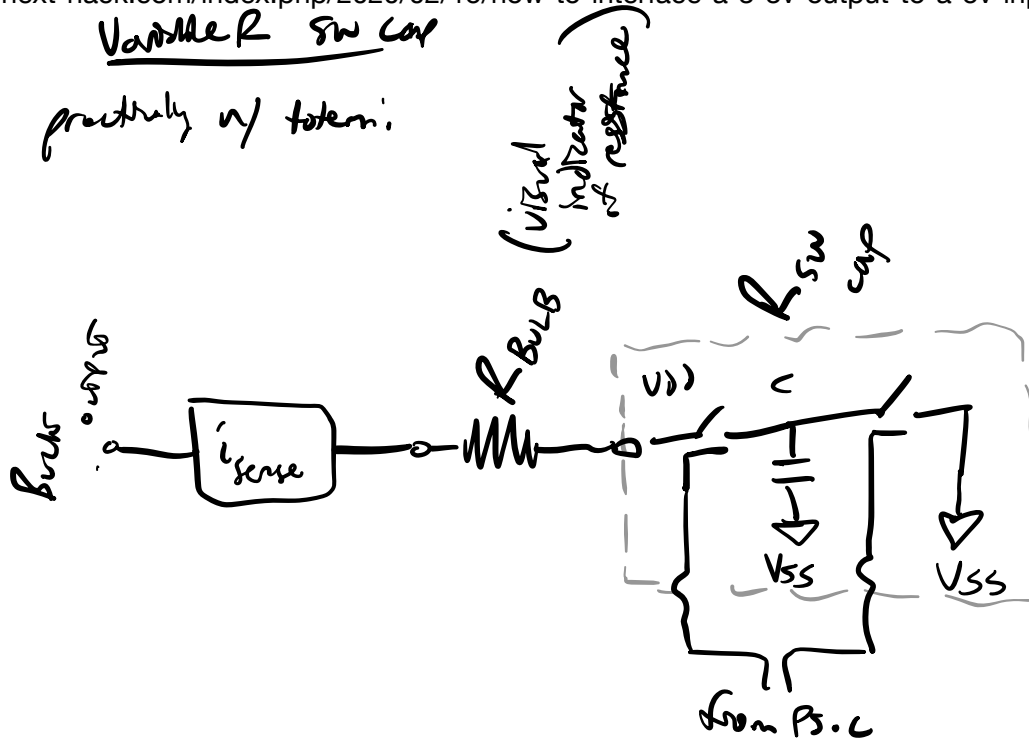
→ P FET (but won't work if $V_{DD} < V_{GS}$)
 → need low side FET for C_b charge



octocoupler typically 80% efficient

and/or non linear...

Variable R sw cap
practically w/ totem.



Analogy in $\rightarrow$ Variable clock PWM $\rightarrow$ digital out (to PSoC)

- TODO:
- schematic for SW CAP VAR R
 - PCB for SW CAP VAR R
 - PSoC and pin $\rightarrow$ clock troubleshooting in C code
- see clock "Set from the PSoC" system

SELF - START CIRCUITRY

FOR CONTROLS

- switched cap?

PSOC SLP

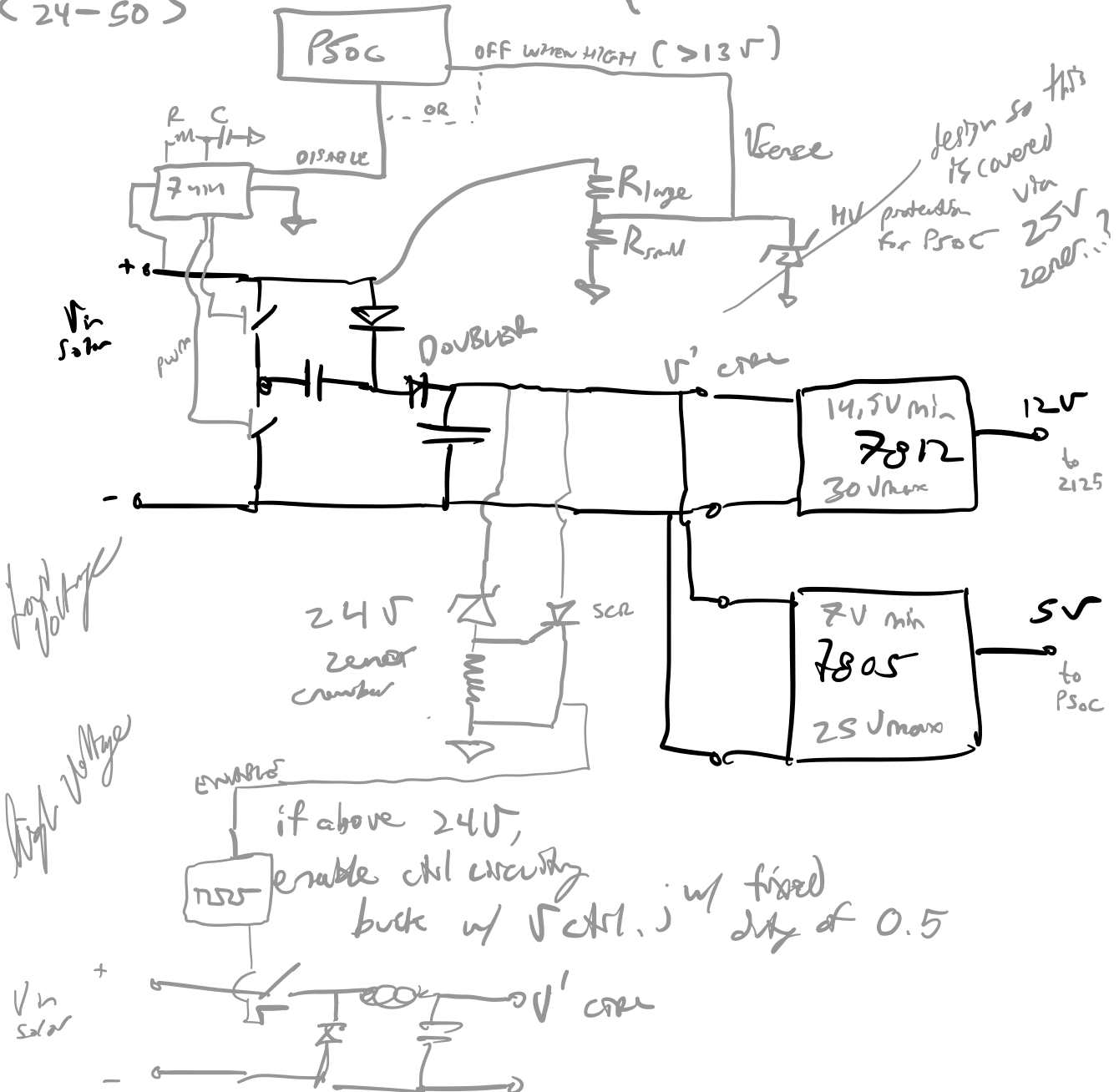
3.8 Boost converter internal... 3.3V
0.5V min

5V
5.5V max

doubler (for startup of diodes?)
12V 2.25?

{ 0-13
13-24
24-50 }

(when off, $V_{in} = V'$)



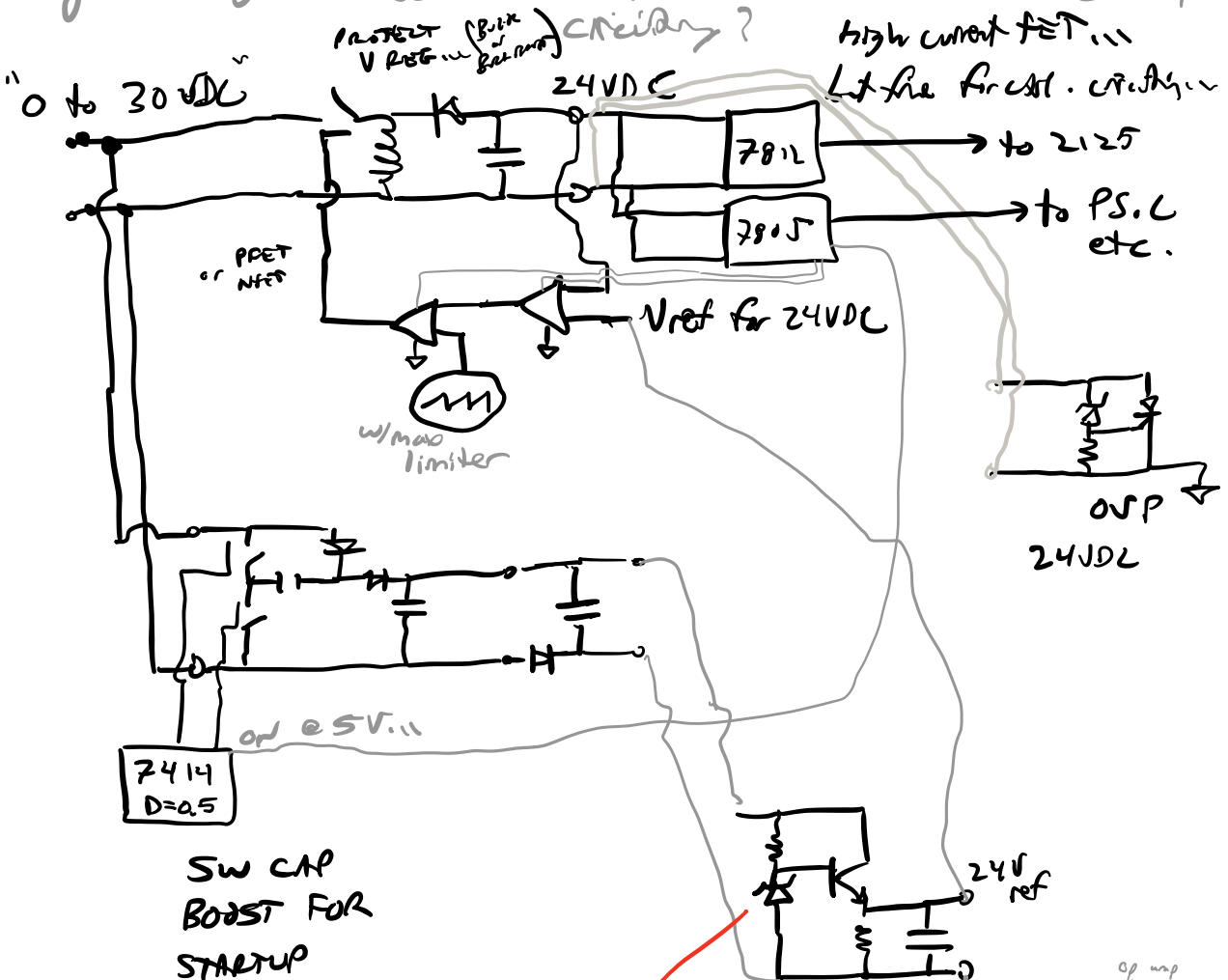
① PSOL on → PWM on → SW cap on from PV

PV high
 ↓
 SW cap off, PWM off
 ↓
 2125 powered from PV panel

← lin. regulator to 12V for 2125

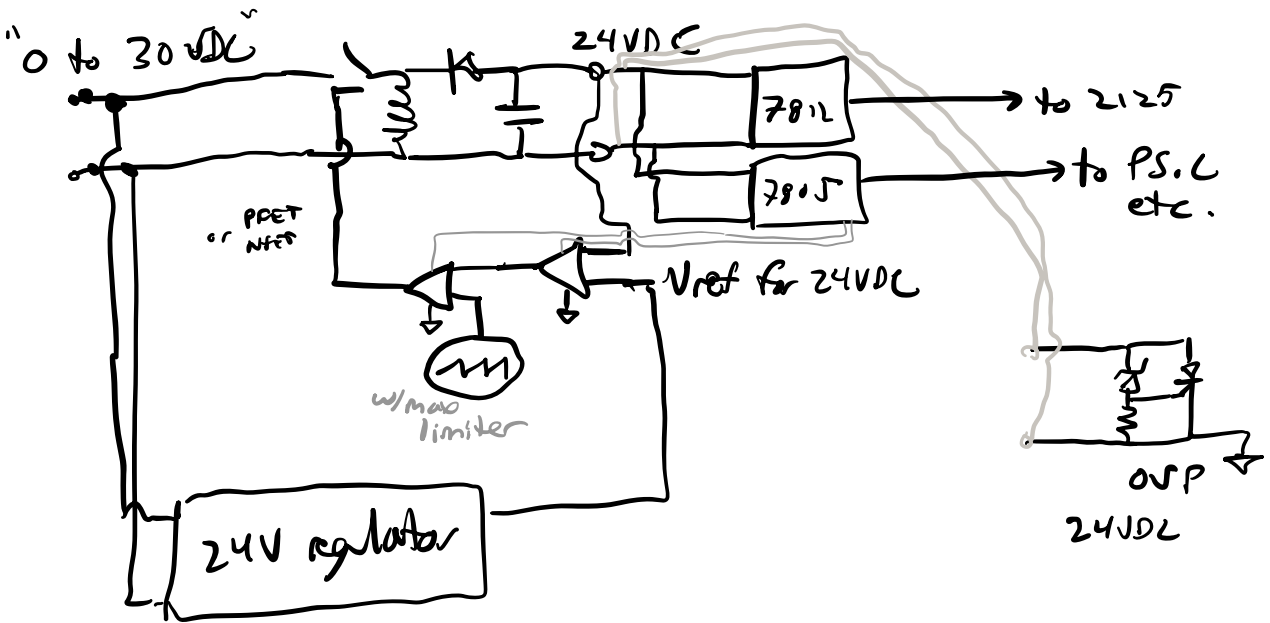
Buck Boost CTL. CIRCUITRY power supply

Why not just controller buck-boost for ctrl startup

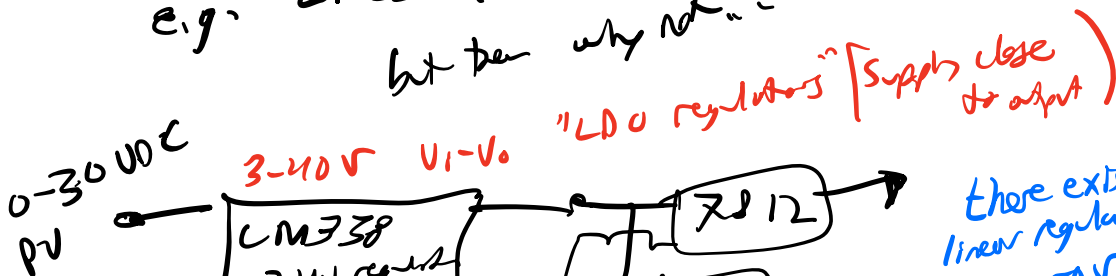


(-)
 Max $I(V_{app})$
 before break...
 Max back diode voltage

op amp limited by supply rails...
 or... just let turn on when 7V min...?



e.g. LM338 ...
but then why not ...

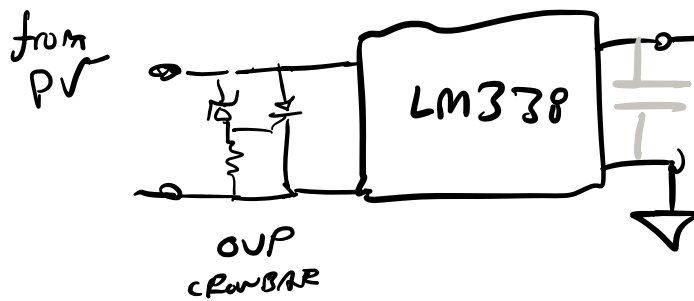
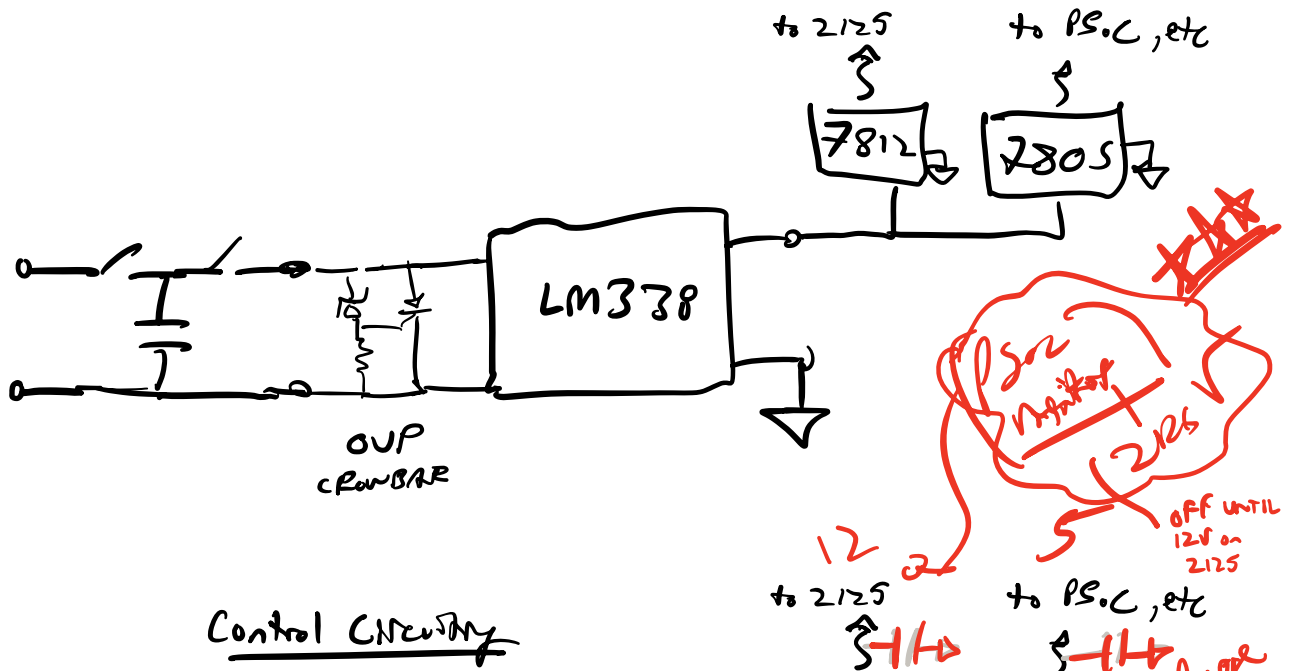


maybe
can't handle
high V? ...
↓ 2? ...

for ctrl.
circuitry? ...

there exist
linear regulators up
to 450V input
w/ 5V
output ...

op amp
limited by
supply
rails ...
or ...
just let
turn on
when 7V
min ...



Power ph,
monitor
OVP
UVP



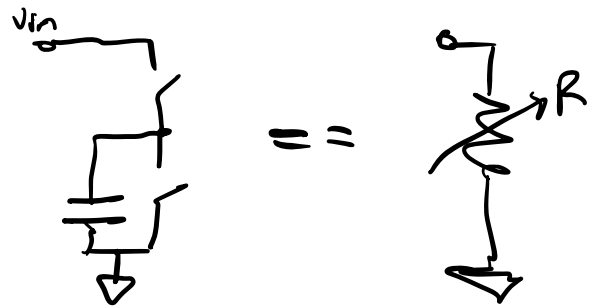
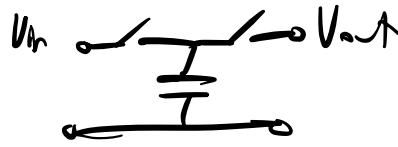
60V
Components
withstand
(see if ur)

I can use sw. cap as a resistor
programmable by f

$$R_{eq} = \frac{1}{C_s f}$$

↑
from P306

greater resolution

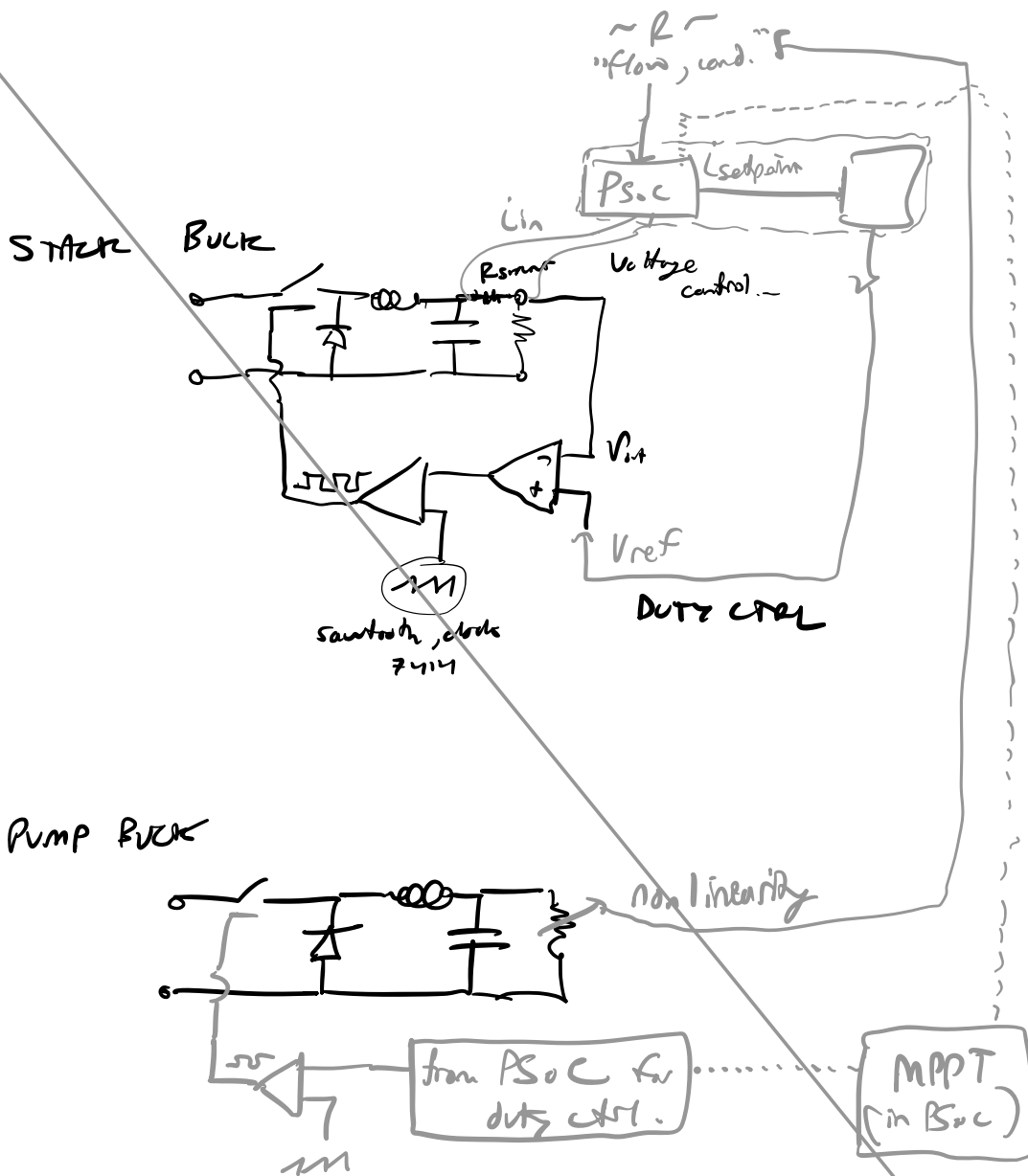


Which can simulate
the electrolyser stack!

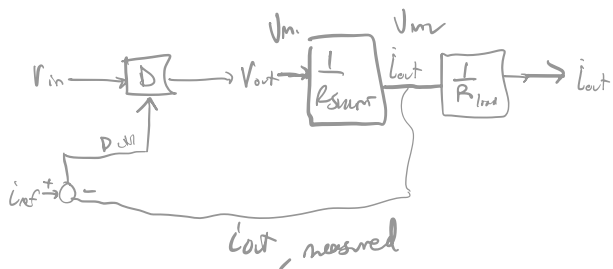
clips { 36V — 60V — 85V — 100V }
Si siming 8

Hydro Power Control

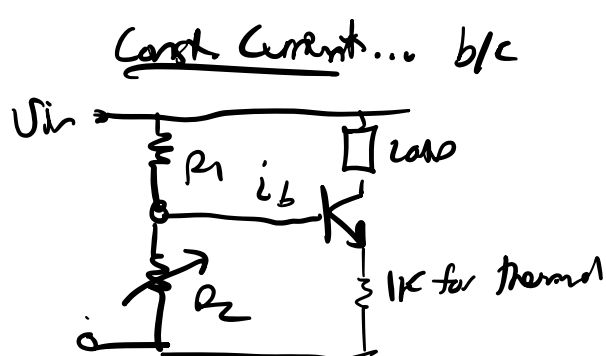
No...



CONTROL



§



$$i_c = (\beta + 1) i_b$$

base voltage

$$I_L = \frac{V_b - 0.6}{R_e}$$

But $\rightarrow$ bad for thermal dissipation

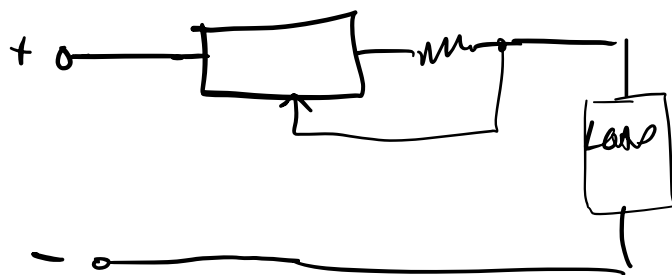
Vary R_2 to ctrl. current...

• const $i \rightarrow$ LED, batteries...

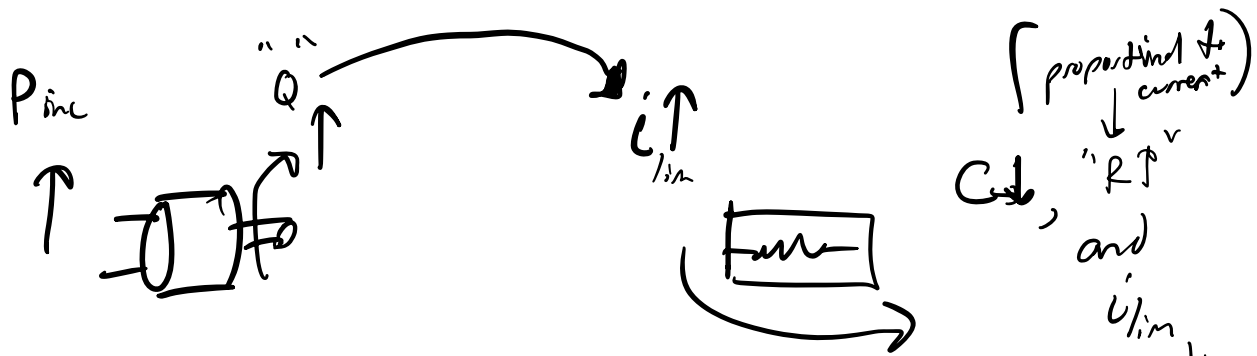
§ transient voltage suppressor diode



§



§ why not just dly control?..



(proportional to current)
 $\downarrow$
 $C \downarrow$, " $R \uparrow$ "
 and
 i_{lim}
 $\downarrow$
 eventually

$$i_{lim} f(C, Q)$$

$$f(R, \omega)$$

$$f(R, v)$$

$$Q \uparrow \rightarrow i_{lim} \uparrow \rightarrow P \uparrow \rightarrow AC \downarrow$$

(less resistance
 time -
 but more
 current -)

$$f(Q, C)$$

$\uparrow$

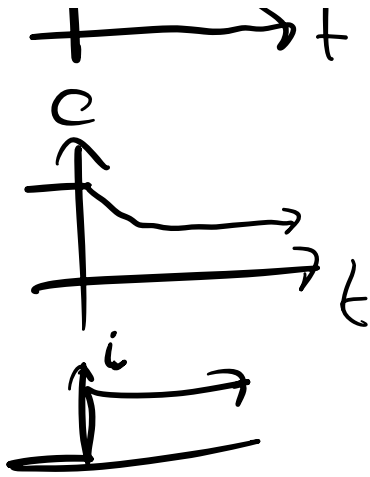
$\downarrow$ until balanced

less res. T
 dominates,
 over
 current
 benefit...

so

Coil, at wire
 stability





allowing more i_{lim} ,

but then at a certain point, i_{lim} so high that $C_{d12, out}$ drops, balancing i_{lim} .

Page 196 - Sun May 28 11:11 AM 2018

Electronics Tutorials

Bypass Diodes are used in parallel with either a single or a number of photovoltaic solar cells to prevent the current(s) flowing from good, well exposed to sunlight solar cells overheating and burning out weaker or partially shaded solar cells by providing a current path around the bad cell. Blocking diodes are used differently than bypass diodes.

Bypass diodes in solar panels are connected in "parallel" with a photovoltaic cell or panel to shunt the current around it, whereas blocking diodes are connected in "series" with the PV panels to prevent current flowing back into them. Blocking diodes are therefore different than bypass diodes although in most cases the diode is physically the same, but they are installed differently and serve a different purpose. Consider our photovoltaic solar array below.

Bypass Diodes in Photovoltaic Arrays

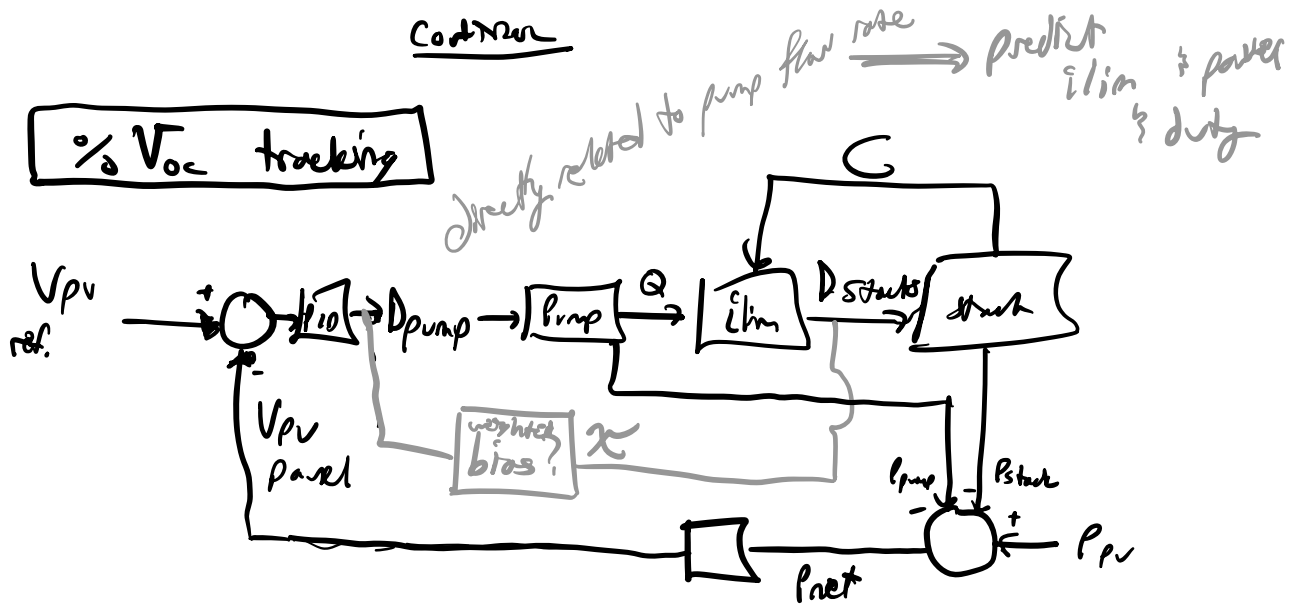
As we said earlier, diodes are devices that allow current to flow in one direction only. The diodes coloured green above are "bypass diodes", one in parallel with each solar panel to provide a low resistance path. Bypass diodes in solar panels and arrays need to be able to safely carry this short circuit current.

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Cookie Settings Accept All

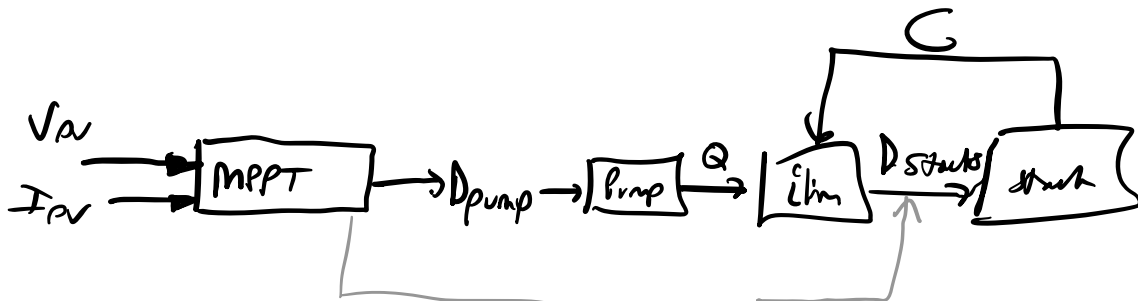
Control

% V_{oc} tracking



Retire and observe

quick ramp down, slow ramp up?
(more sensitive to over- than under-shoot)



respond quickly for power down
prot threshold...

PV: 12-21V & duty

RULES / GLANCE

nonlin: P_{Soc}: — if driver $V < 12VDC$

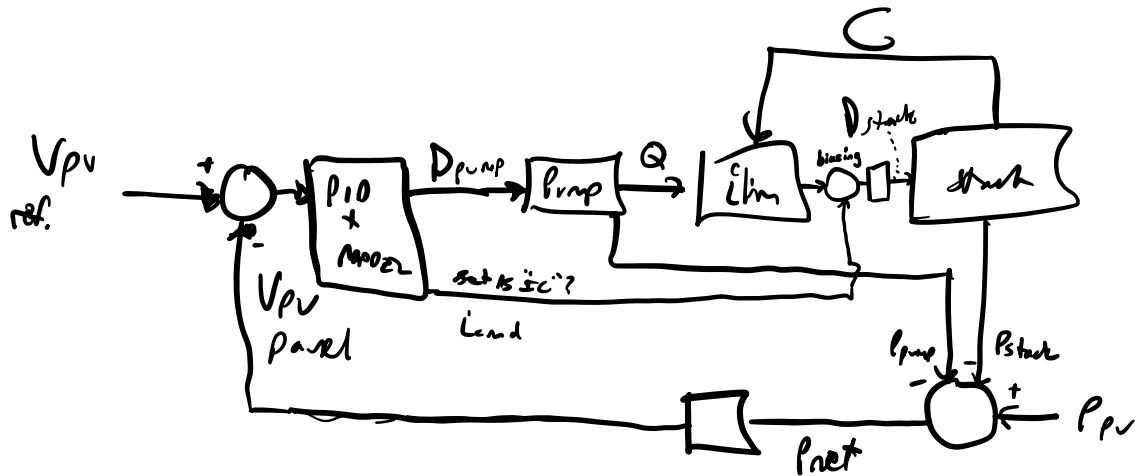
FETs control 0 DUTY

— if PV power drops 'rapidly'
or threshold ($< 15VDC$ PV, OFF)

Control

% V_{oc} Tracking MODEL-INFORMED
INITIAL CONDITION

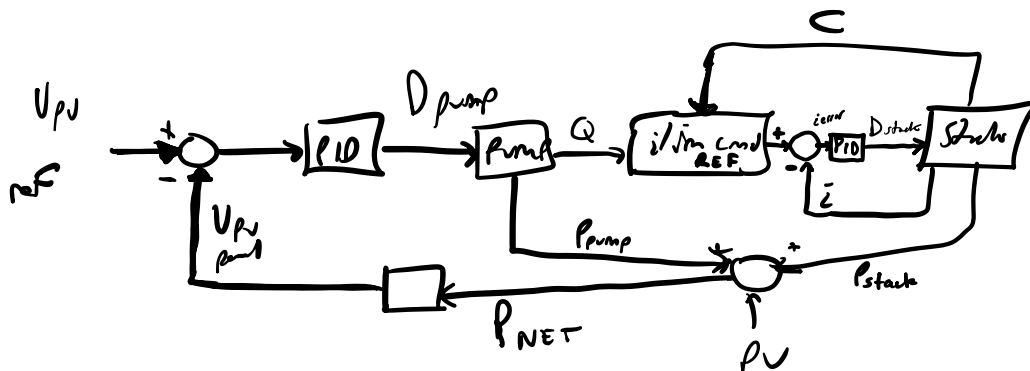
(eh...)



pre-emptive lookup table
Directly related to pump flow rate $\Rightarrow$ predict i_{lim} & power & duty

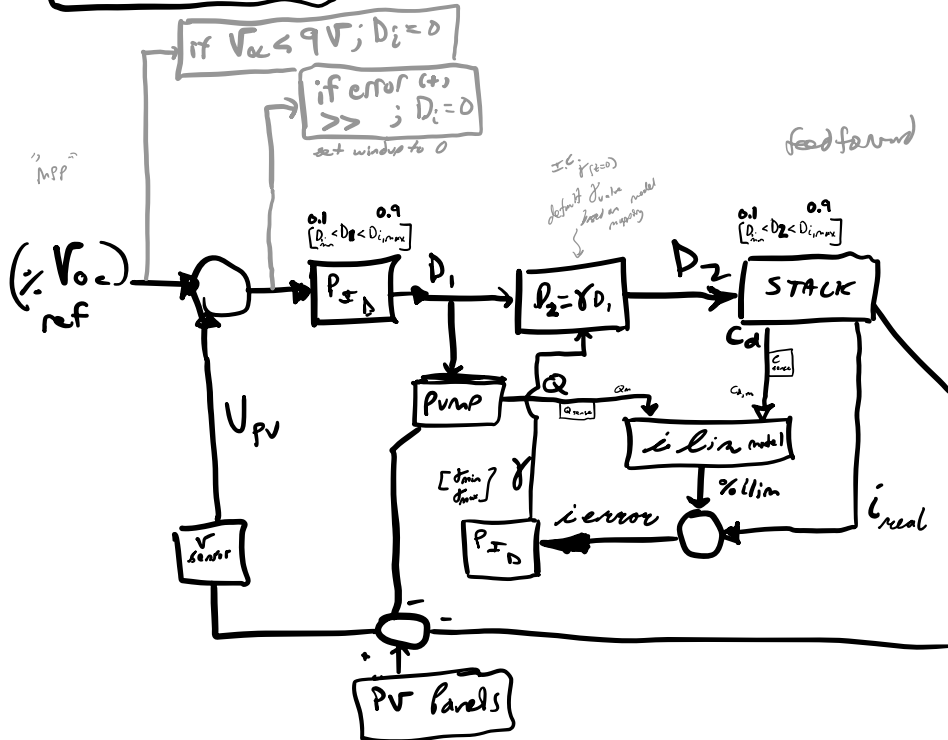
% V_{oc} tracking CASCADE

(closest to post)



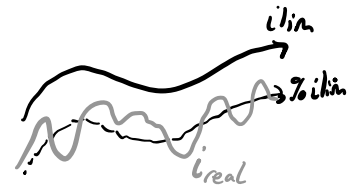
% V_{oc} Tracking CASCADE w/ LENIANCY

(lenient)

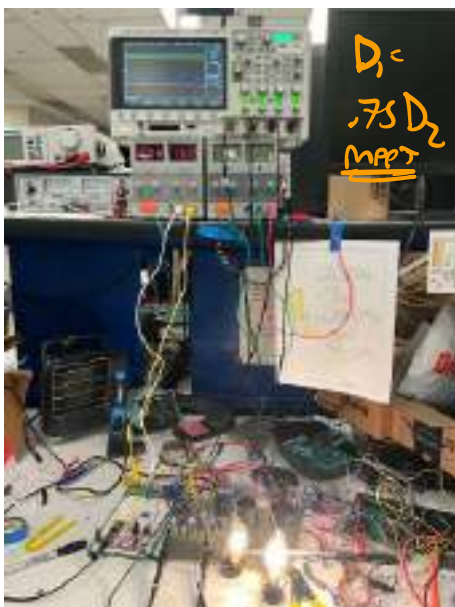


γ changes based on i vs. i_{lim}

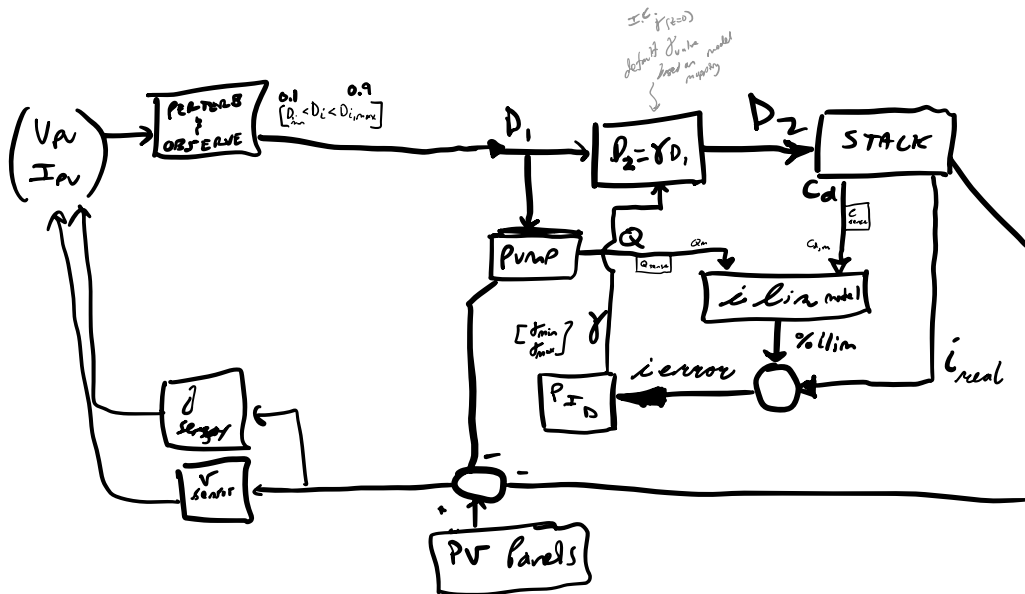
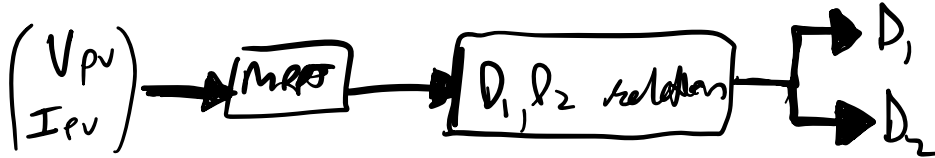
here, i need not be exactly i_{lim}



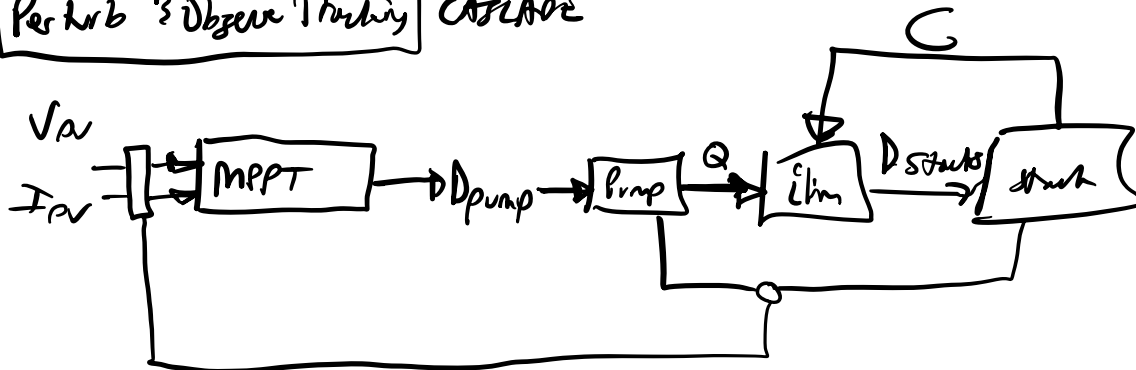
allows leniency to avoid power overdraws



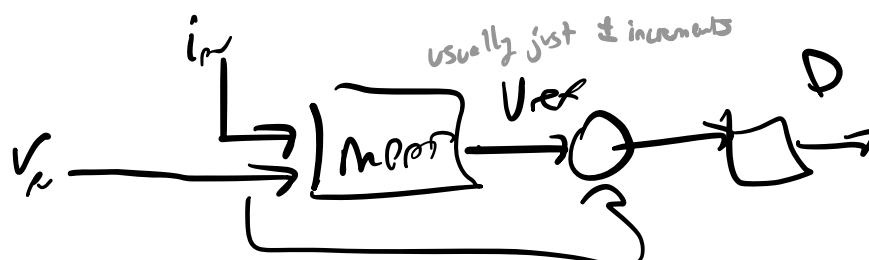
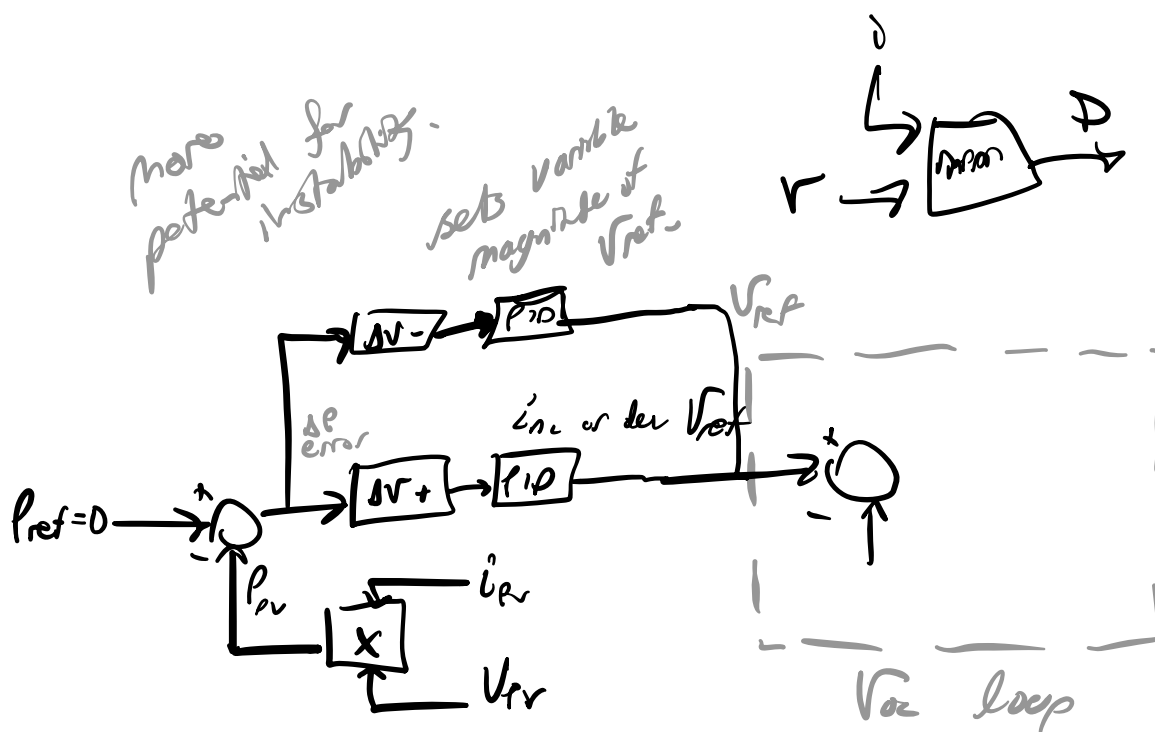
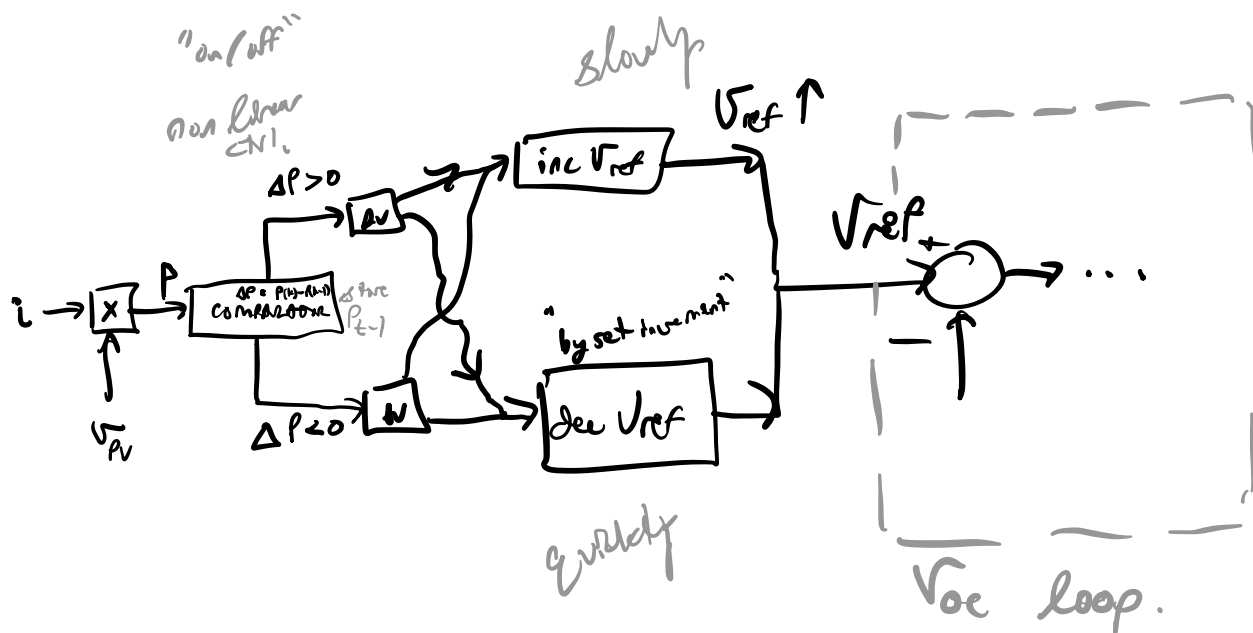
Two point
power comparison
method



Perkeb. & Usaha Tani **CASLAVE**



PID } "difficult"
SMC } to design



Pto
→ reasonable dynamics

limits / drawbacks
→ power oscillations

↔
perturbation
step size...

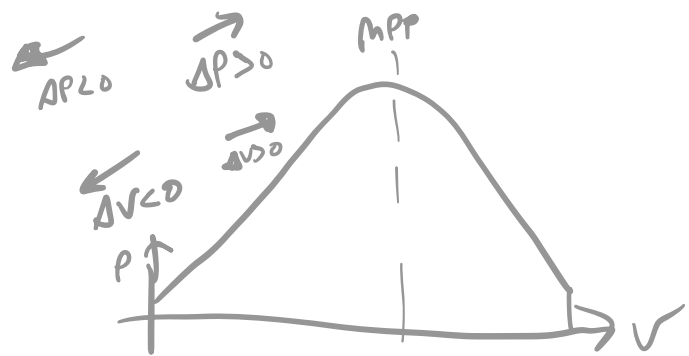
↓
smaller reduces oscillations S.S.
BUT

bad N. tral. response.

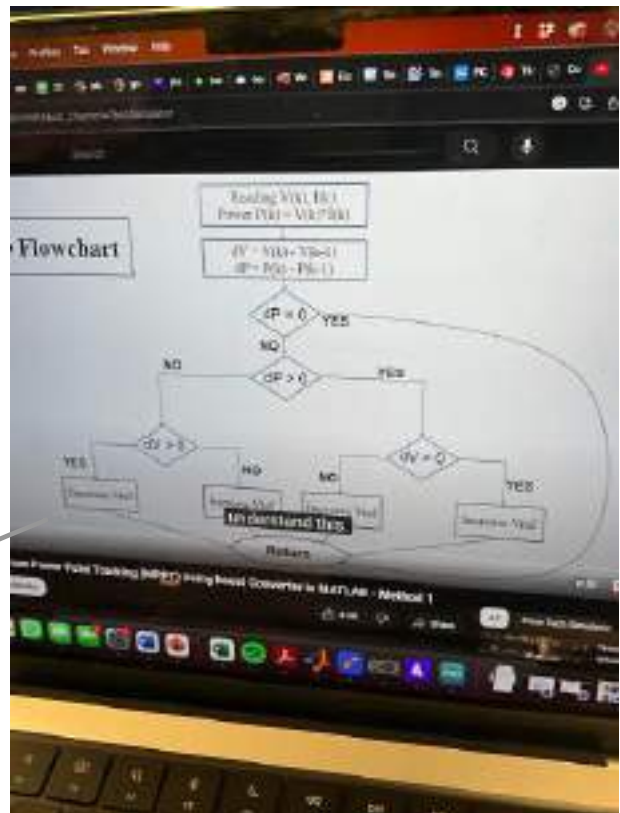
→  local min...

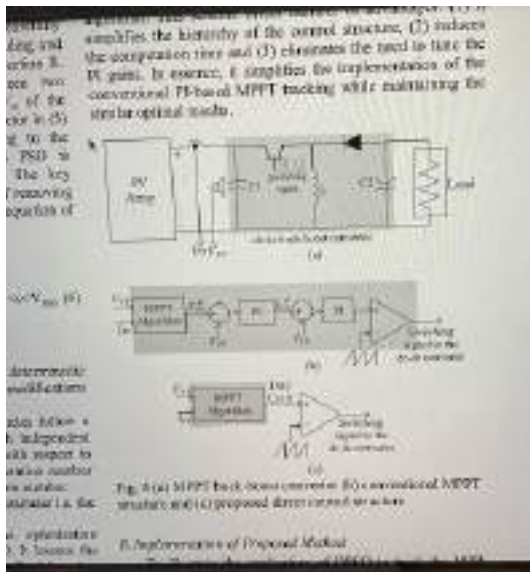
→ "drifting"

(inc
D
dec etc.)



← → ← →
 When $DV < 0$ When $DV > 0$ When $DV < 0$ When $DV > 0$
 and $AP < 0$ $DP > 0$ $DP > 0$ $AP < 0$
 inc V inc V dec V dec V



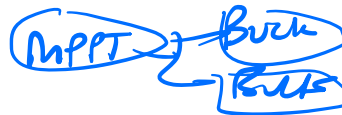
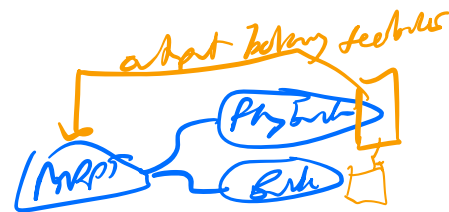


Ishaq 2012
Salam

For TLD: CTRL arch

IEEE Arch

H Bridge?



STEPS TO BUILDING

→ V_{dc} w/ 2 ; V_{dc} w/ 2 bricks

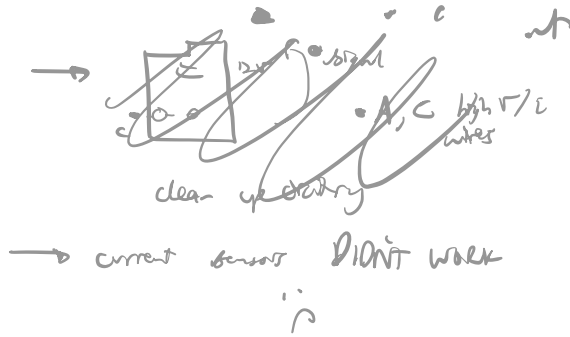
→ P_{30} w/ 1 brick ; P_{90} w/ 2 bricks

P_{50} w/ 2 bricks & biasing

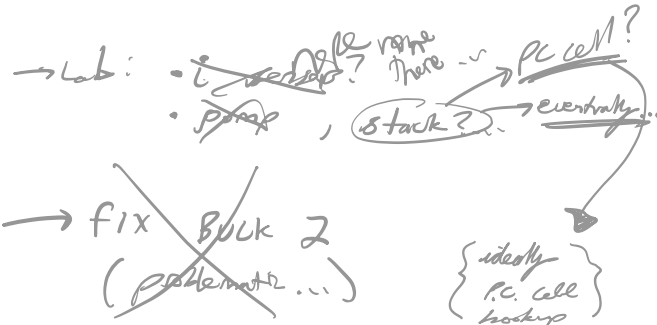
IEEE Transactions on Power Electronics



12/6/23



→ MPPT ^{P&O} 1 buck algorithm, ...
2 hr



12/7/23

⇒ Fix & CHARACTERIZE PSC var R

C	S	R
■	⋮	⋮
■	⋮	⋮



⇒ i sensor, P&O algorithm

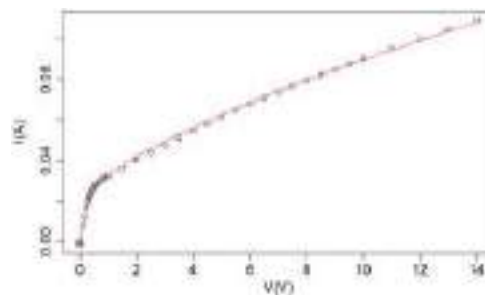
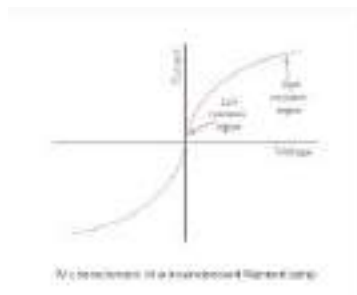
12/8/23

⇒ P&O alg. tuning params.
2 load P&O (lights vs. resistors?)

delay? (PSOL)
and/or
math board

⇒ implement $\sigma(\text{mA})$ rim current measurement function
 ↳ need sensor N, size, etc.
 ↳ 35,000 rpm ilim? depends on
 cm/s
 (density gradient of size...)

⇒ implement cond, Q sensing



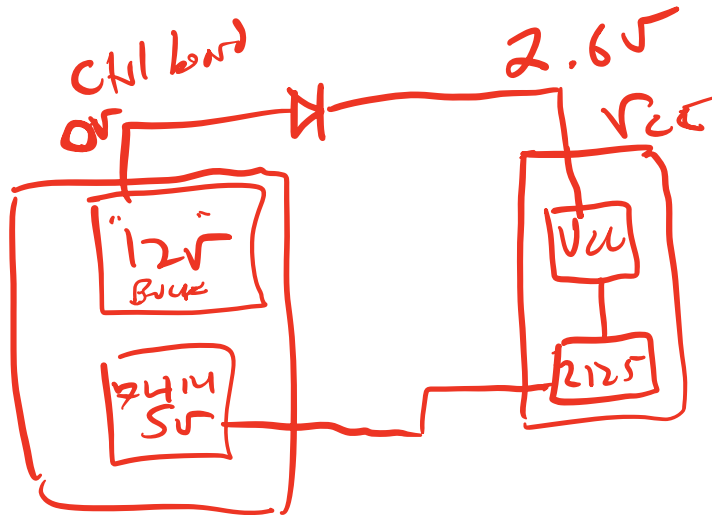
5V from 7414 ⇒ causing $V_{ce} \frac{1}{2} 5V$
 $\sim 2.6V$

5V $\rightarrow$

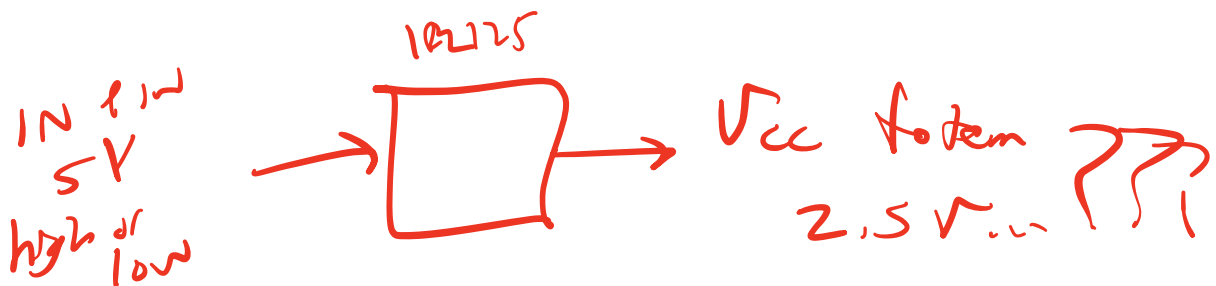
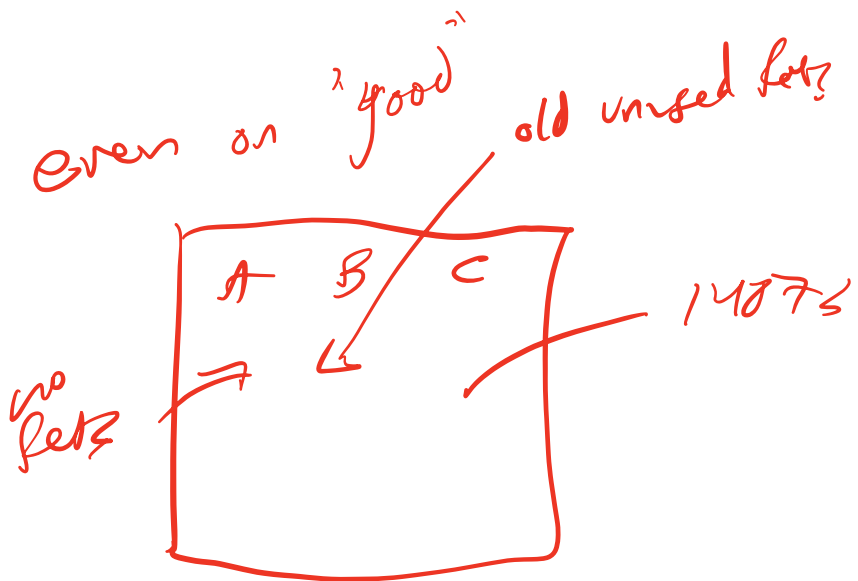
1	8
2	7
3	6
4	5

(ISOLATED
AT DRIVER
CHIP 1R2125)

even w/ diode $\rightarrow$ not working



When 2125
out, good in



GROUNDING ISSUES?

on same board
when GND, fms

~~Linear regulator dead!~~

~~red 120ohm low power buck~~

~~5V~~

~~1 7805W-500~~

~~5V, 2.5W~~

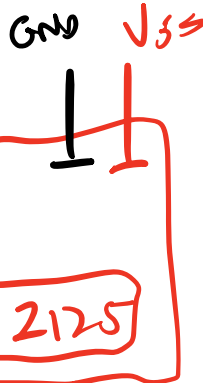
~~500mA
9V → 7.2V input~~

Nu, when on separate P.S.

GND, or o.c. on V_{SS}
 V_{CC}

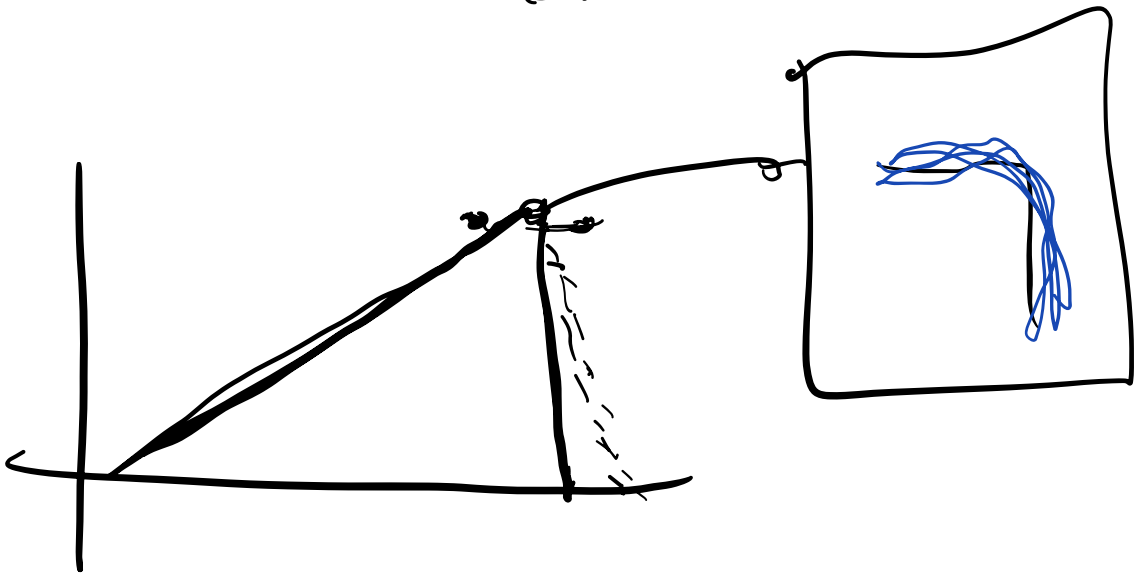
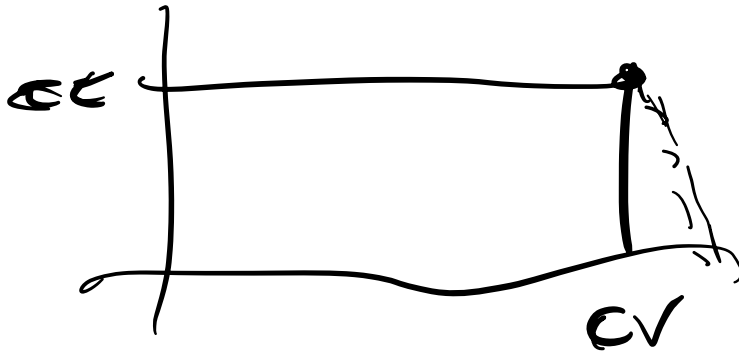
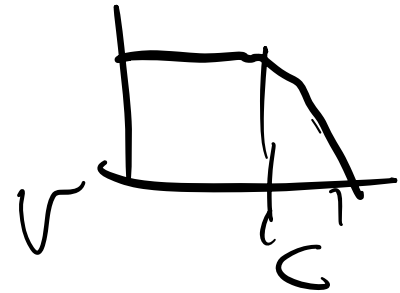
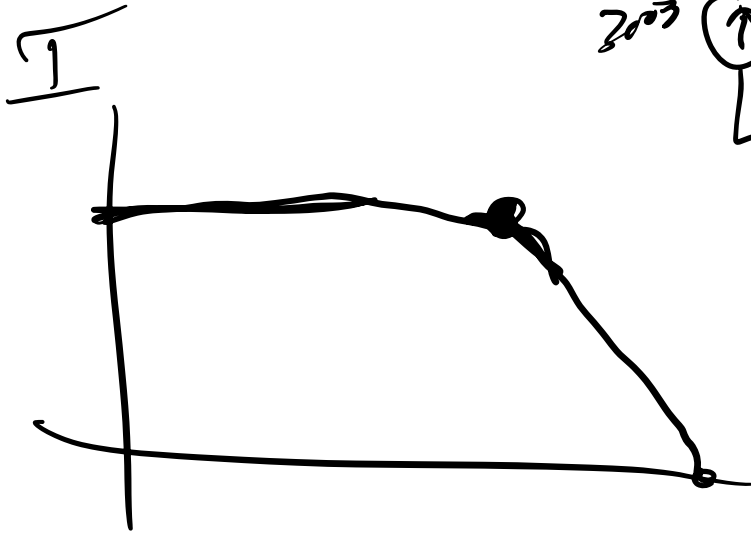
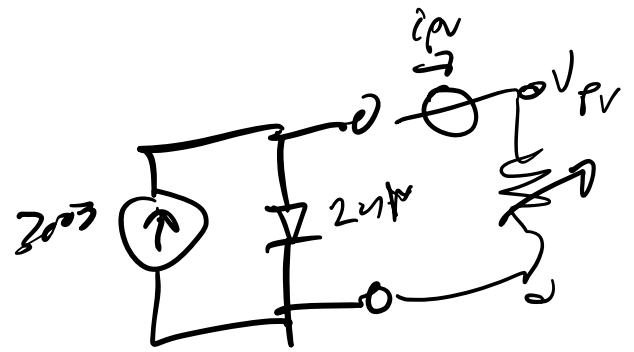
$\left(\begin{array}{l} 9V \rightarrow 3V \\ 8V \rightarrow 2.6V \end{array} \right)$

$V = \sim \frac{1}{2} V_{signal}$



V_{signal}

$\left\{ \begin{array}{l} 32/9/23 \rightarrow 2N25 \text{ on A, H} \\ \text{fried multiple times in FET?} \\ \text{Unknown case,} \\ \text{mally to CHB} \end{array} \right\}$



$$cmd_{min} = (duty \times SB)$$

$\nearrow$ $0-256$
 $\nwarrow$ integer
 $\swarrow$ score factor
 $\searrow$

$$\left[\text{round}\left(\frac{0-256}{256}\right) \times (2) \right]$$

$$[0-2]$$

$$cmd_{max} = duty \times SB$$

$$\stackrel{=}{256}$$

$$\frac{256}{256} \Rightarrow 1$$

$$\frac{256}{128} \Rightarrow 2$$

$$\left(\frac{cmd_{max}}{duty} \right)$$

$\downarrow$
 0.9

$$\times SB \text{ side factor} = \underline{\underline{max \text{ size}}}$$

i_{lim}

Wait until
i settled ~30s

4 min/dez
1 min/test

$C \times L$	$Q^0 [mL/min]$	V	i
300	?	?	?
600			

$C \times L$	$Q^0 [mL/min]$	V	i
700	?	?	?
1200			
1500			

$C=500$

5 cell pairs
Unleak spacers
— cm^2 mem area
— cm^2 anode area
electrode type

Walt

$C \times L$	$Q^0 [mL/min]$	V	i
300			
600			

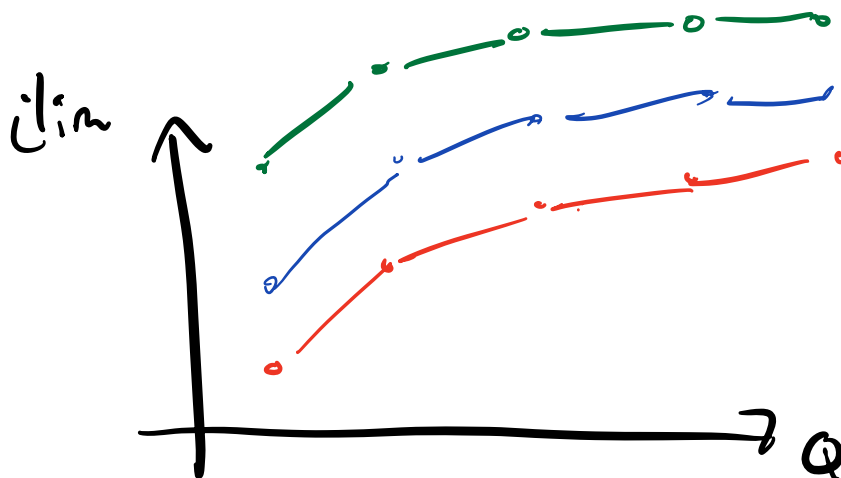
$C \times L$	$Q^0 [mL/min]$	V	i
700			
1200			
1500			

$C \times L$	$Q^0 [mL/min]$	V	i
300			
600			

$C \times L$	$Q^0 [mL/min]$	V	i
700			
1200			
1500			

$C=1000$

$C=2000$



@ $C=500$
 $C=1000$
 $C=2000$

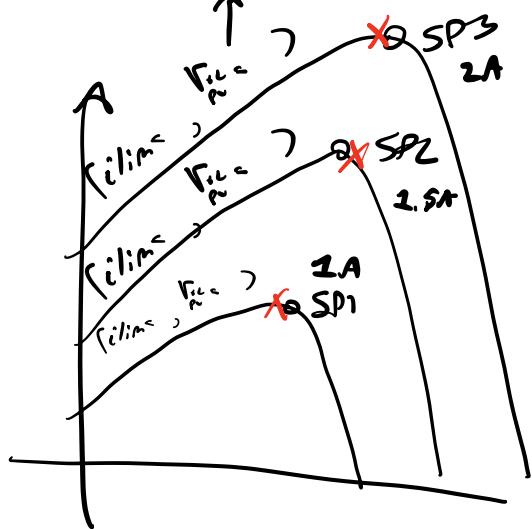
$$i_{lim} \propto C Q^B \quad \text{fix } \alpha, B$$

PAPER FOR 6.210 Paper

STEP RESPONSE

2.8A for 3003 power supply
 $5.11A I_{sc}$
 (20.4V nominal)
 24V_{oc}

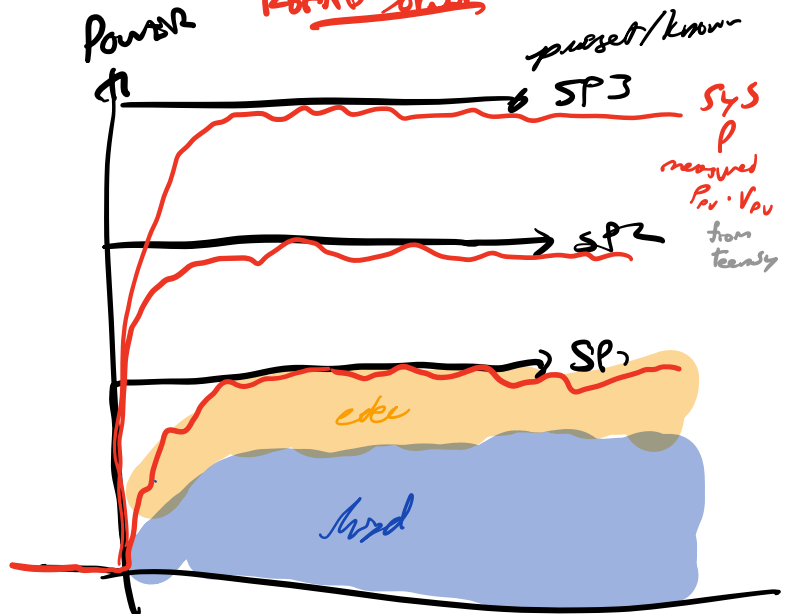
meas. w/ multimeter (how)



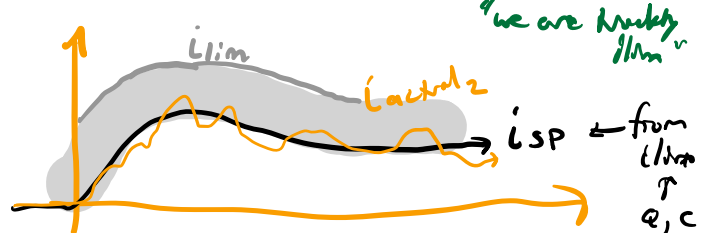
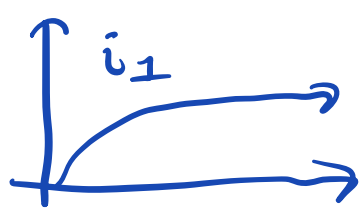
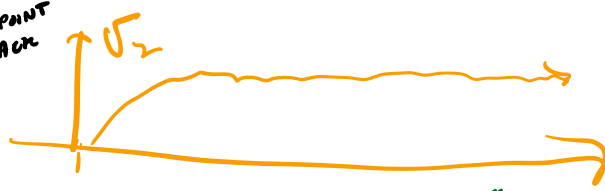
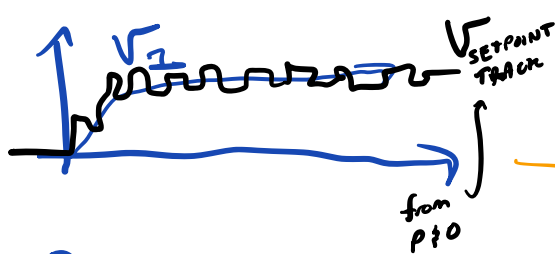
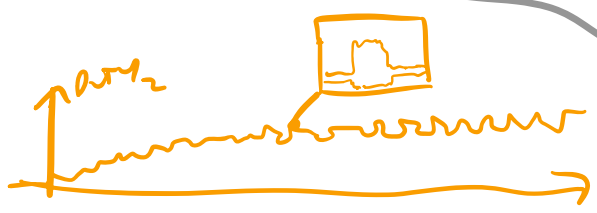
"we are tracking MPP"

X₁

(PRESET)
 3003 off, TRIEN on
 KOFAD 30.3A



from Teensy



Q T ~ ~ ~ ~ ~

"if cloud"
 maybe intermediate step down
 (how)
 for wind max
 → P_{actual}

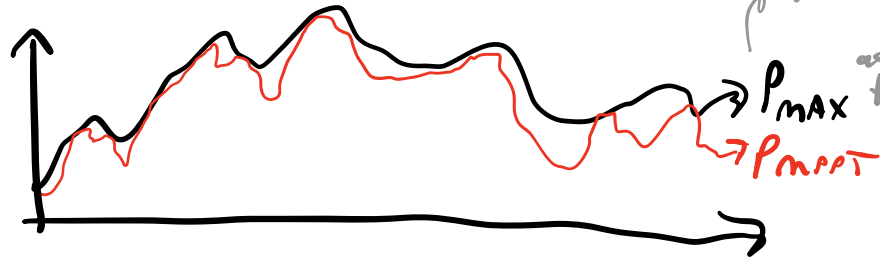


time-wise

"we track over wide profile"

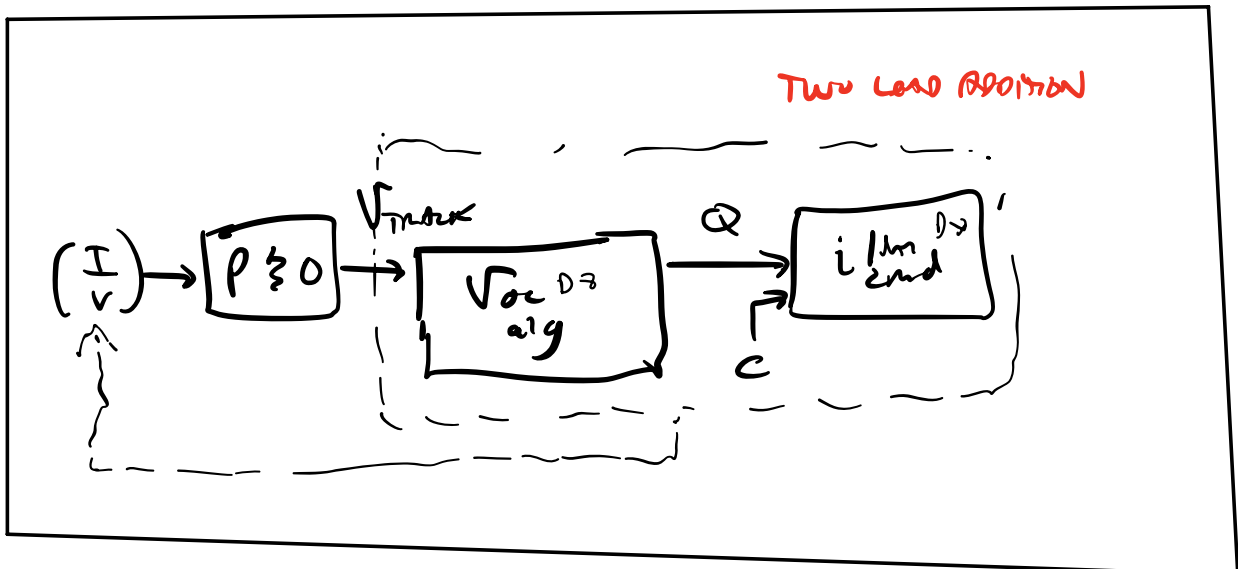
(how)
 drive from
 → i → 0.05?

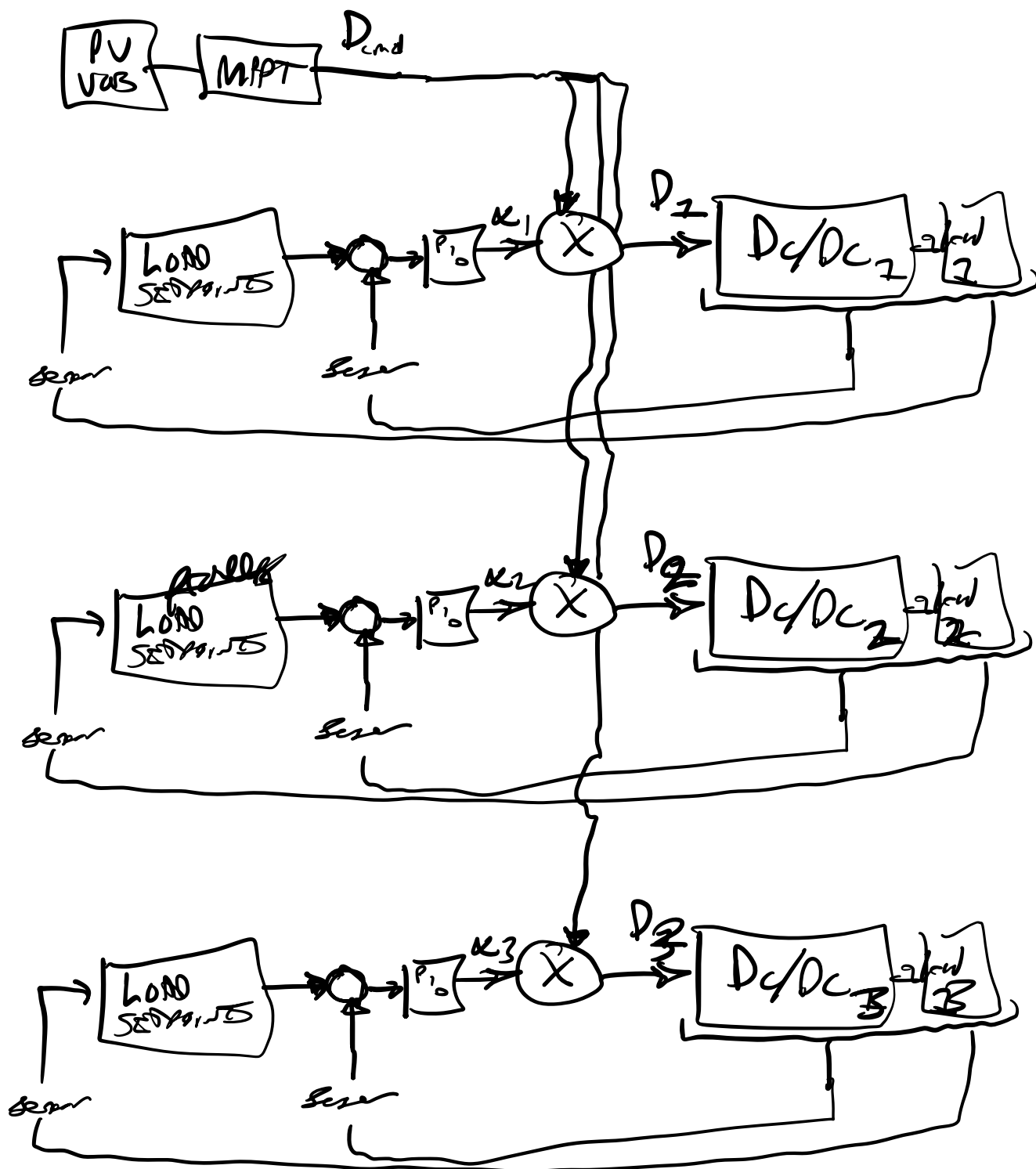
R_2

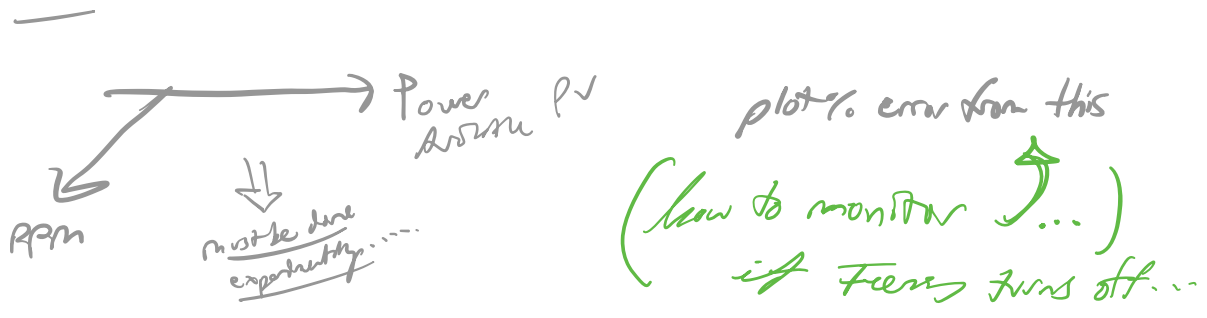


as I turn i knob...
 → P_{max}
 → P_{maxPT}

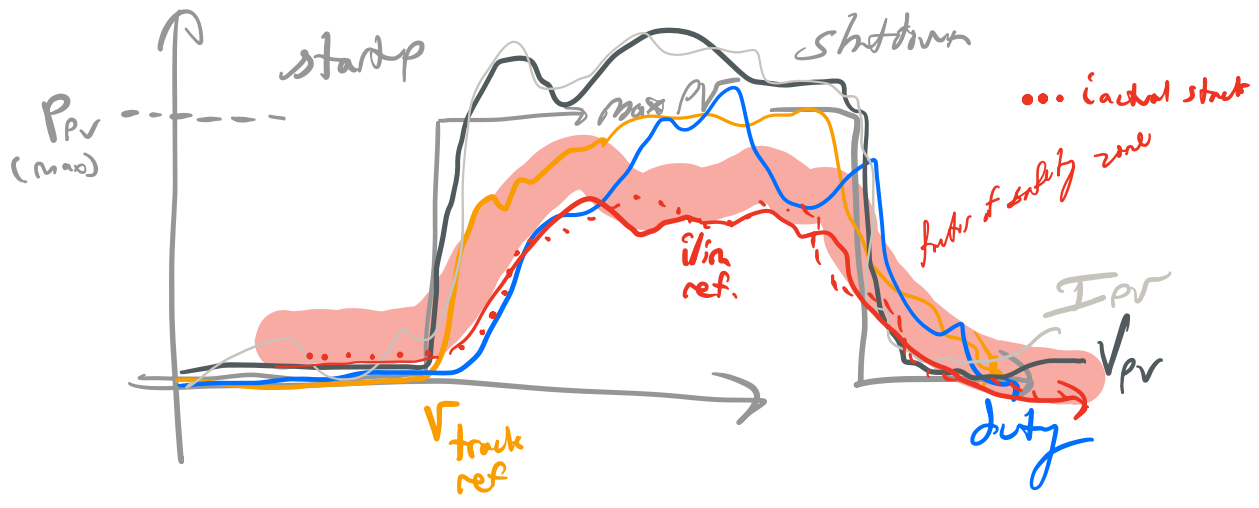
- f₁₀
 params
- f_s



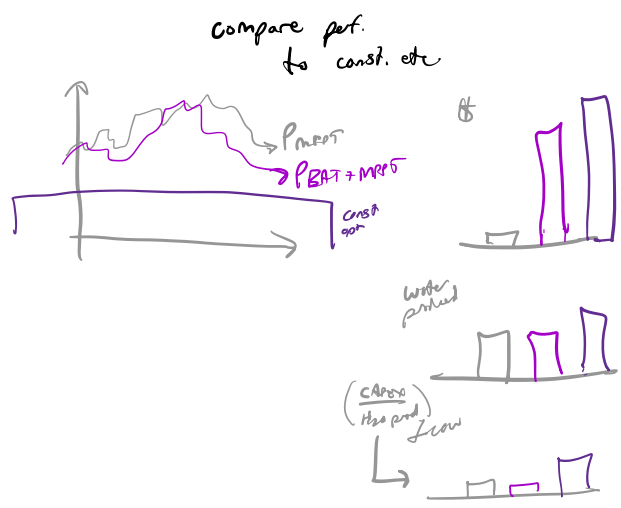
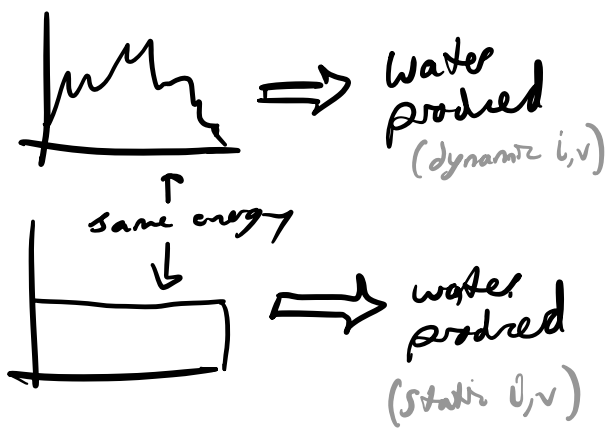




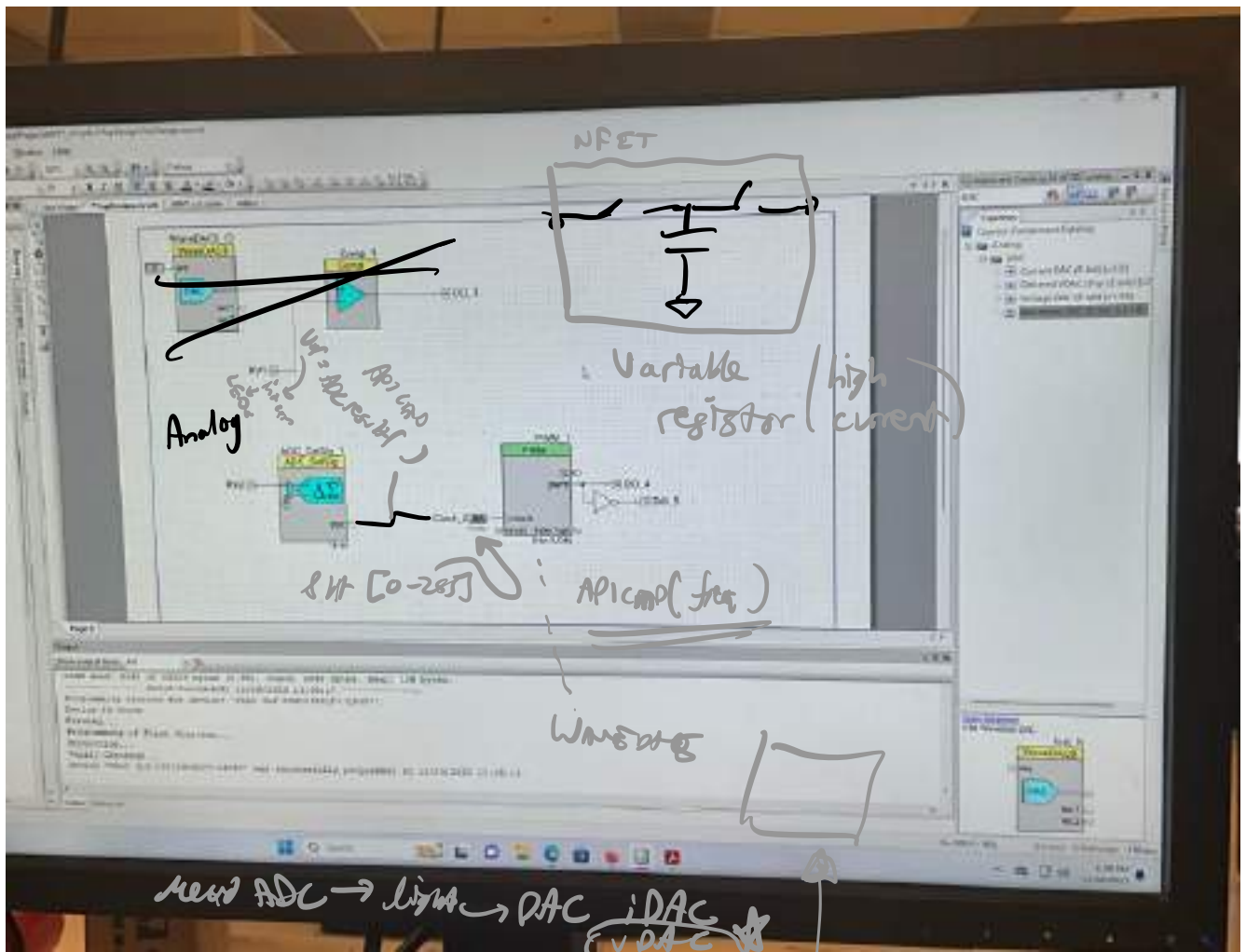
• STEP RESPONSE (for a given tuning)



• Methods / metrics to compare ctrl to state of the art?



comparable H₂O produced w/o batteries,



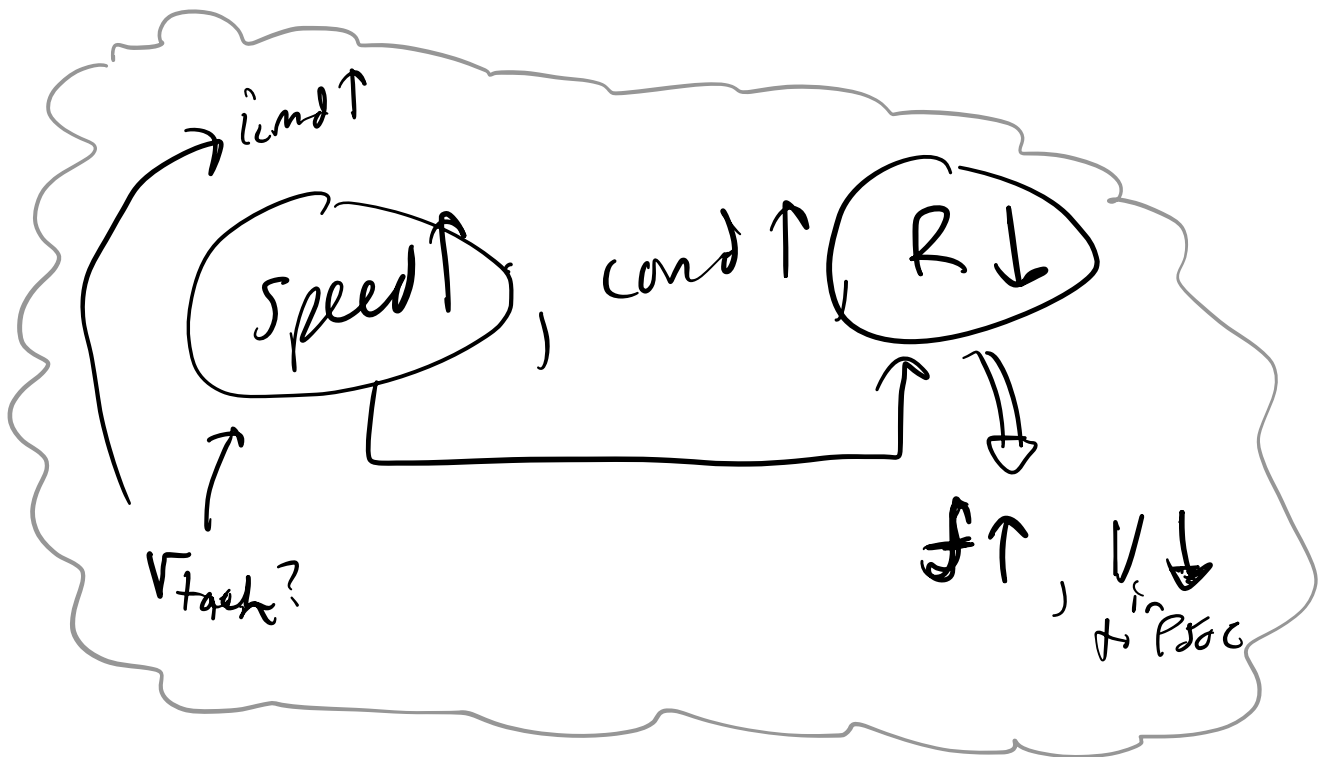
new ADC → DAC → DAC
 VCO
 ↓
 { voltage controlled oscillator }
 freq. to voltage converter...

direct memory access
 ADC → DMA → Timer → Waveform

Proc lecture



Clock	8 bit Pwm	Analog V	Digital 32 bit	Mapped Value	Meas f
100 kHz	390 Hz	2.25 V	2,147,483,650	32768	8.2 Hz
					3.05 Hz
100 kHz	390 Hz	0.98 V	841,813,590	12846	7.78 Hz
					100 Hz
		0.11 V			858 Hz
		4.75 V			3 Hz
		.23			500 Hz



$$f_{desired} \cdot 256 = f_{clock}$$

$$\frac{1}{\left(\frac{\left(\frac{1}{f_{desired}} \right)}{256} \right)} = f_{clock \text{ required}}$$

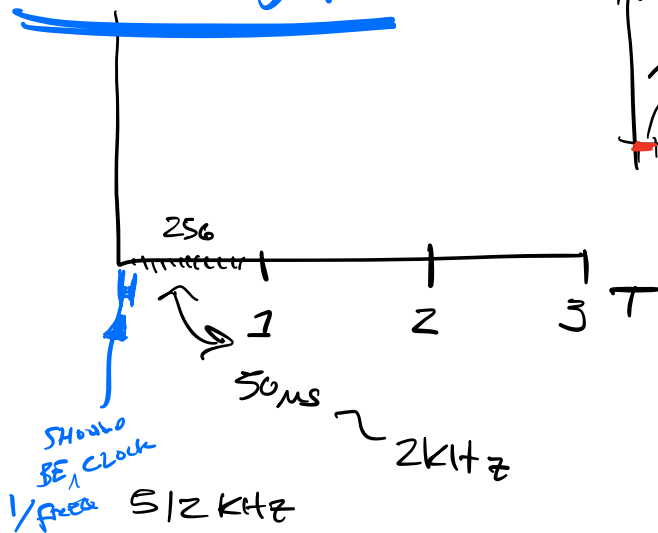
↑ bits

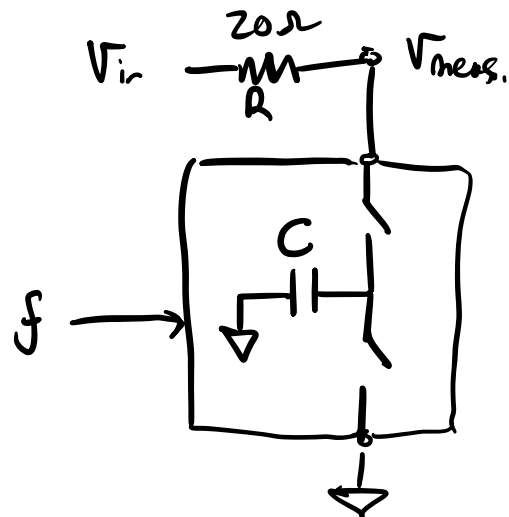
$$\left(\frac{1}{2 \text{ kHz}} \right) = T \dots 1/256 = T_{clock}$$

↓

$$f_{clock} = f_{clock}$$

notes on setting PWM



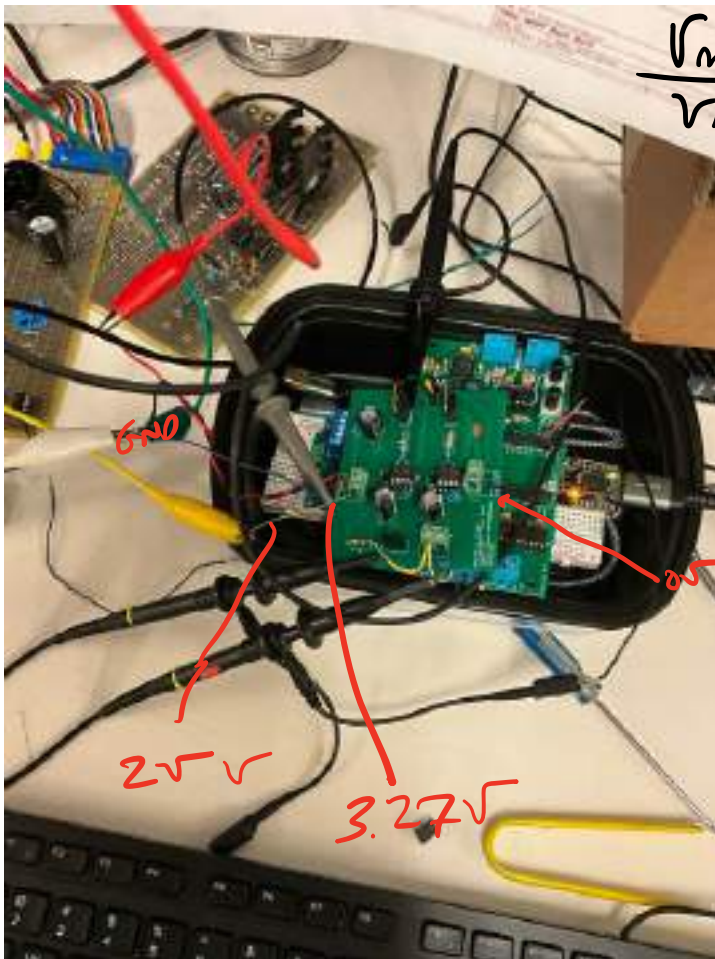


$$V_{\text{meas}} = V_{\text{in}} \left(\frac{R_2}{R_2 + R_1} \right)$$

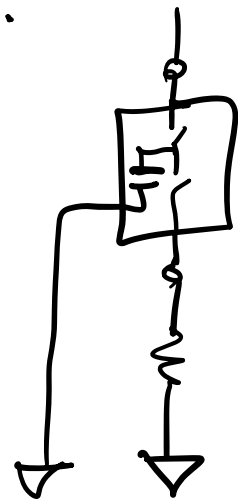
$$V_{\text{meas}} R_2 + V_{\text{meas}} R_1 = V_{\text{in}} R_2$$

$$V_{\text{meas}} R_1 = (V_{\text{in}} - V_{\text{meas}}) R_2$$

$$\frac{V_{\text{meas}} R_1}{V_{\text{in}} - V_{\text{meas}}} = R_2$$



try:

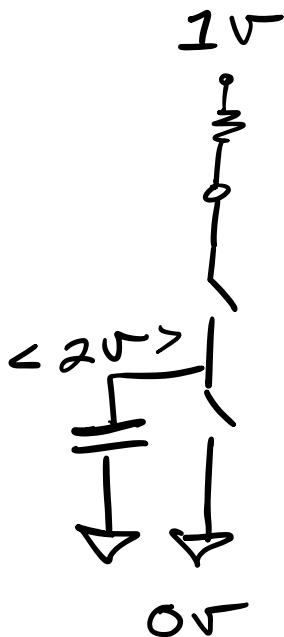
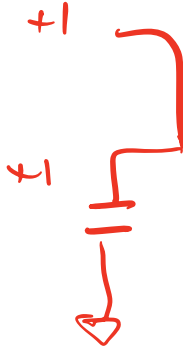


during, this is a
 $V \uparrow$ (but $i \downarrow$)

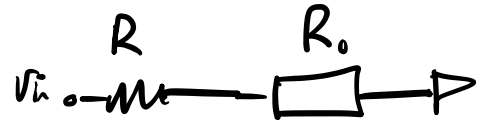
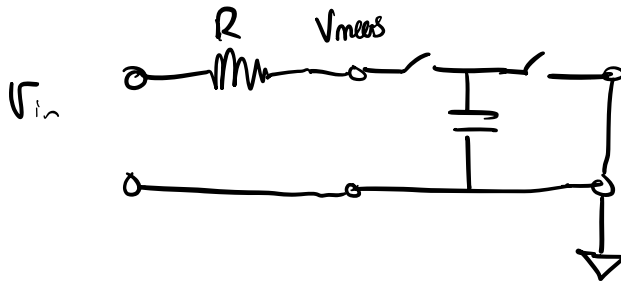
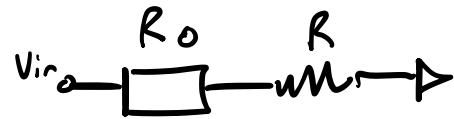
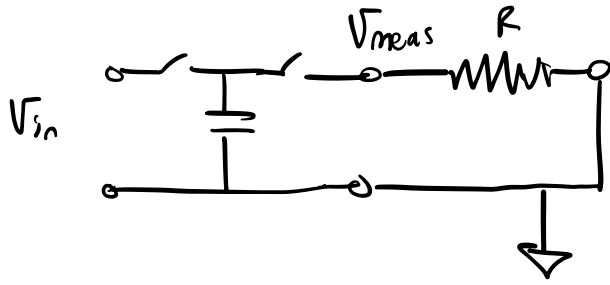
perhaps voltage divider doesn't apply

would
not
change
voltage...

but should
consume power?



$$R_0 \sim \frac{1}{Cf}$$



$$V_{\text{meas}} > V_{\text{in}}$$

e.g. @ 10 Hz, $C = 0.1 \mu\text{F}$
 $D = 0.5$, $R = 10 \text{ k}\Omega$

$$V_{\text{in}} = \begin{matrix} 1 \\ 2 \\ 3 \\ 4 \\ 5 \end{matrix}$$

$$V_{\text{meas}} = \begin{matrix} 2.14 \\ 3.08 \\ 3.89 \\ 4.65 \\ 5.41 \end{matrix}$$

When I expect

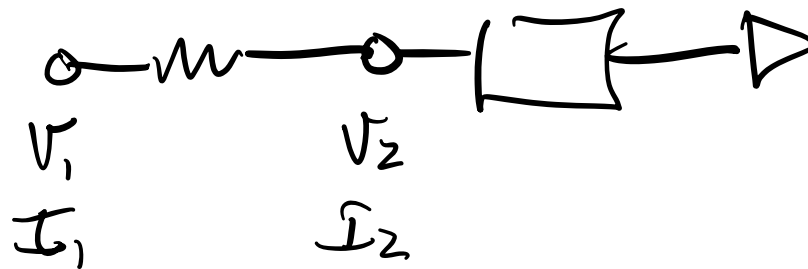
$$R_0 = \frac{1}{Cf} = \frac{1}{(0.1 \times 10^{-6})(10)} = 1 \text{ M}\Omega$$

so,

$$V_{\text{meas expected}} = V_{\text{in}} \frac{R_0}{R + R_0} = (1) \left(\frac{1 \text{ M}\Omega}{10 \text{ k}\Omega + 1 \text{ M}\Omega} \right) \approx 99 V_{\text{in}}$$

"buffer" / variable resistor...

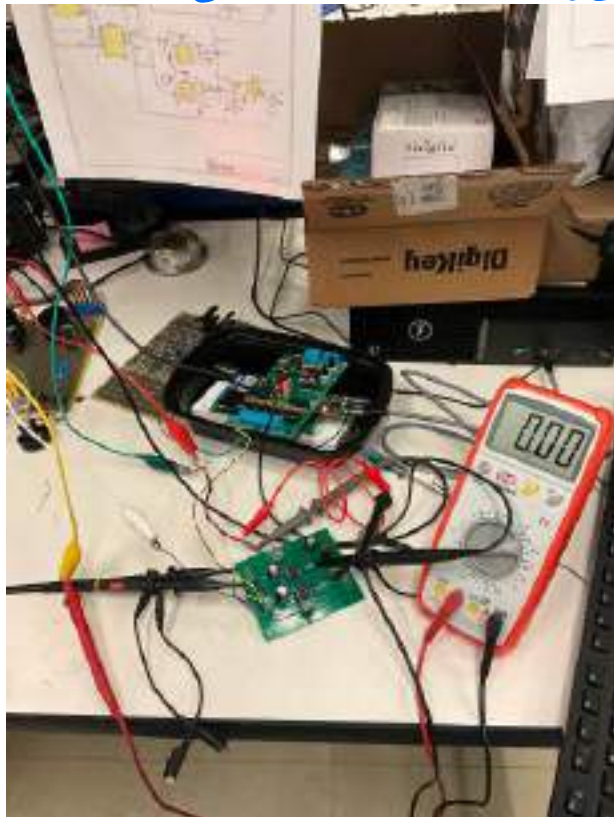
$$P_{in} = P_{out}$$



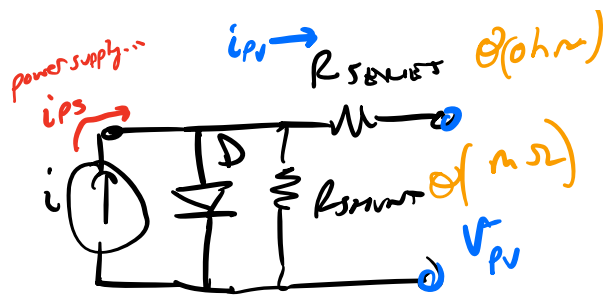
$$I_1 = I_2 \dots \text{perhaps it changes w/ freq?}$$

same current w/ and w/o sw cap.

sw cap var R (exp)



PV model



$$i_{sc} R_s \ll V_{oc}$$

$$i_{sc} \sim \frac{30 \text{ mA}}{\text{cm}^2}$$

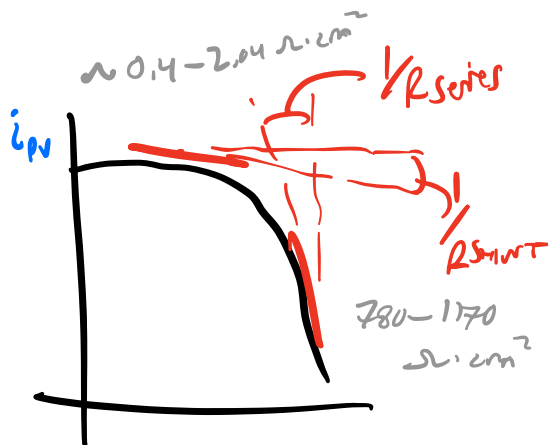
$$V_{oc} \approx 0.6 \text{ V}$$

$$R_s \ll 20 \Omega \cdot \text{cm}^2$$

Typically $R_{shunt} \sim 100 \Omega \cdot \text{cm}^2$

100 W Panel ; 18V
6A @ 12VDC

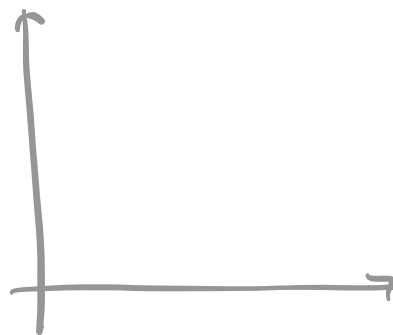
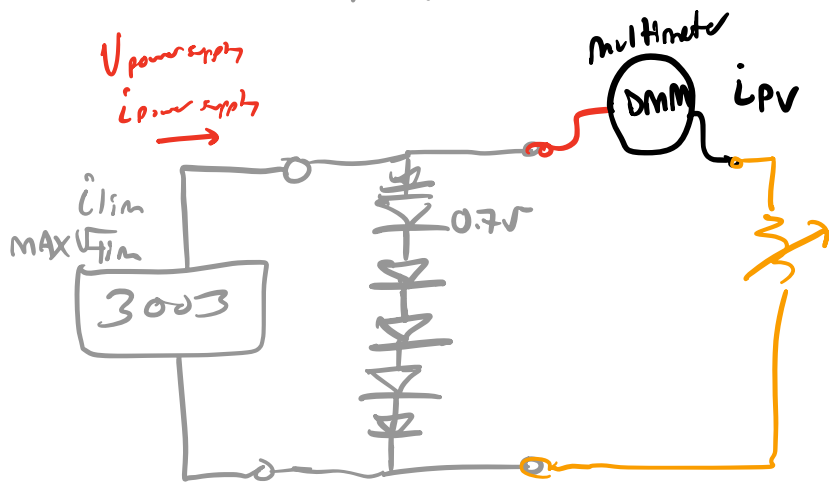
6839 cm²



[2016 Niassé et al.]

Optimization of Electrical
Parameters CdS / iZnO Thin ...

e.g., w/ a potentiometer
and multimeter



Ⓢ $R_{full \text{ s.c.}}$, $i=0$, $V=\text{max} \rightarrow i=0$, $V \approx 1.5 \text{ V}$ for ~~diodes~~ ^{x6 D n, regular diodes}

$i = 0.1$
 $P = 6$
 $D = 20$

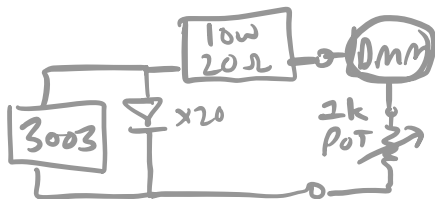
V	Power Supply
5	
14.2	

$D = 20$
 $i = 0.3A$
 lim

PV V	PV i
13.6	0
13.4	0.03
13	0.05
9.9	0.11
0.3	0.12

now $4/ 20\Omega$
 10W
 power R series

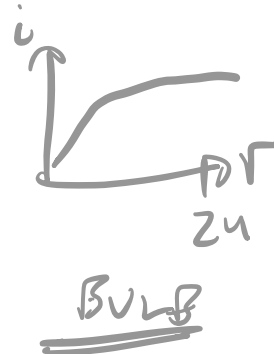
$i_{lim} = 0.10$
 1W POT



PV V	PV I
14.2	0.01
13.9	0.02
13.8	0.03
13.5	0.04
13.2	0.05
12.9	0.06
12.6	0.07
12.5	0.08
12.2	0.09
11.8	0.10
9.9	0.11
4.9	0.12
2.7	0.12 + ...

"24V" supply

lamp
 $\sim 260mA$
 $\sim 24V$
 Max



$0.05 \text{ A} \rightarrow 2.52 \text{ V at ?}$
 inversely mixed perhaps

from 0.05 A
 2.52 raw
 1.72 after delay

$0-5 \text{ V}$
 $0-11 \text{ A ?}$
 $1-0 \text{ A ?}$

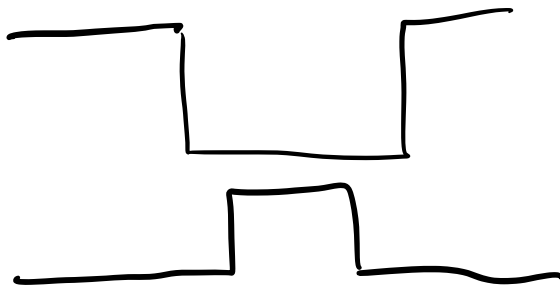
$.450$

45 mV/A

$\rightarrow 0.225 \text{ V expected}$

$4.775 \Rightarrow 3.1515$

~~STM~~
 $2.52, 1.72$

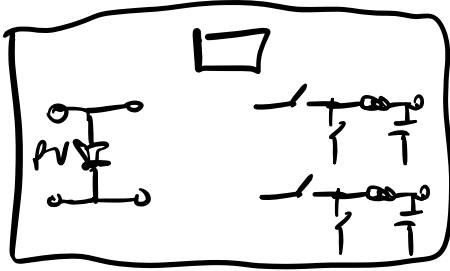


dead time insertion

MVP (12/3/23)

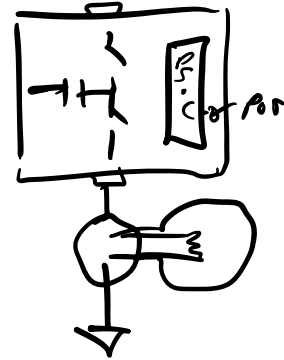
MPPT-BUCK CONVERTER

- data monitoring / visualization
- ports to observe MPPT



• ideally 2 buck MPPT

high current sw. cap
variable load



TO DO:

- calculate expected efficiencies

- buck
- FETs (f_{sw})
- ohmic in ^{losses}
- driver losses?

• V diodes $P = i^2 R$
 $i = V/R$

• CPU draw

• sw cap

• light

TO DO: $P_{OT} \rightarrow f$
 $\downarrow$
expected R out...
(add delay to...)

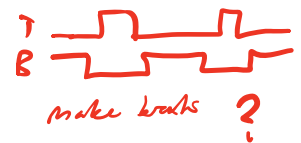
- measure experimentally $\left(\begin{matrix} V_{in} I_{in} \\ V_{out} I_{out} \end{matrix} \right) \} R$

- calc/measure disruption speed?

- show startup/shutdown? -

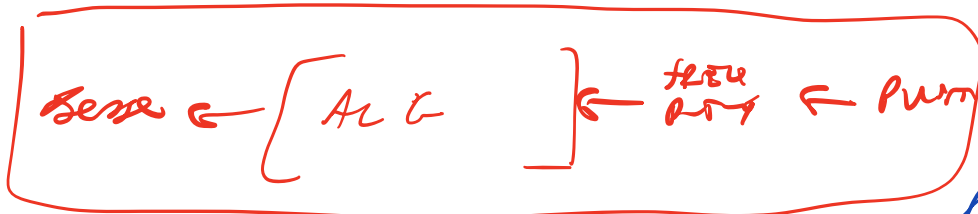
- TL to disclosure $\rightarrow$ MPPT PV BD (ctrl. mps.)

$\rightarrow$ CONT. VOLT LOAD SENSOR



— MPT algorithm formulate!
on Teosy

— validate $\left\{ \begin{array}{l} V \text{ regulators} \\ i \text{ sensors} \\ V \text{ sensors} \end{array} \right\} \left\{ \begin{array}{l} PWM \text{ signal Teosy} \end{array} \right\}$



supervisory ctrl.

for actual dty
(t)

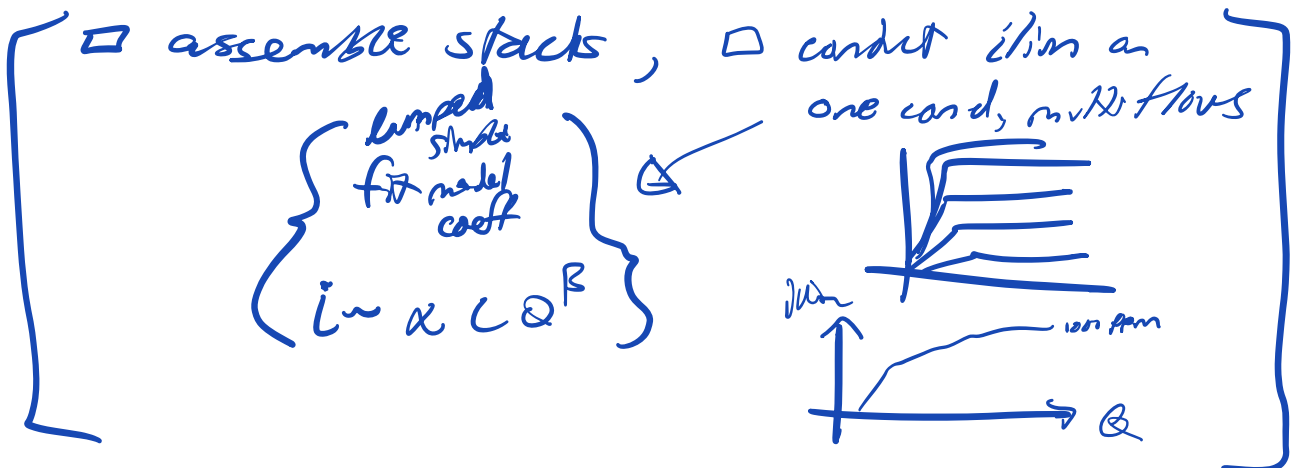
(maybe keypoint system)

— to do: 12/9: sensor recording to

x/x, or
something

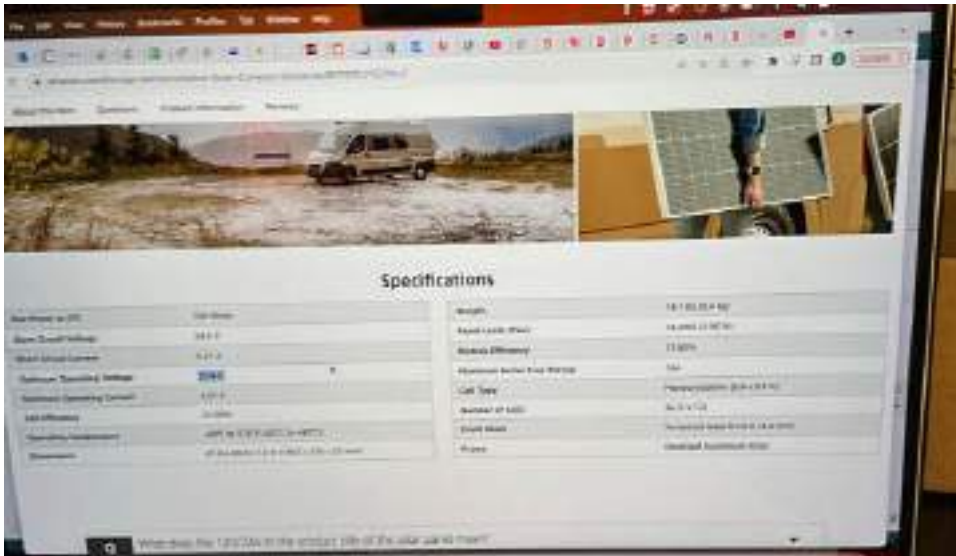
Δi_z sense ΔC sense
 Δi_r ΔQ sense

fix channel A 2125 driver failure



i control alg. PID
w/ setpoint from $x i/lim \leftarrow Q, C$

Kin



- to do 12/12/23

10A ✓ i lin stack coeff.
✓ i calibration

→ Steve OK; ask ✓ to bring in system

12P

• figs

✓ 2125 dydy; current rating on PCB?

(1.2A max per channel)
bt
then 2.4 on Vdd

charge/
discharge

≈ 12-16V
0 (100 mA)

0.54 mm spaces
0.65 mm between

5CP
0.6 void frac
0.45 are pretty

8.201 $\Rightarrow$ tune i sensor /
recalibrate so new Δp
is most accurate

$\Rightarrow$ see i lim response?

$\rightarrow$ try for a step.

if not done by ILP $\rightarrow$

call it / use current results

$\rightarrow$ tune i cmd PID (k_p, τ_i)

1P

6.2224 Totem
• LAB → Card #2 (backup)
→ PSoC variable Resistances...

• $\frac{1}{2}V_{oc}$ record mis() step
for different V_{in}
targets
(show chip settings)

• MRP7 funky.

⇒ cut#up:
PQ → intro
→ histograms Black
CPL
etc.

→ method?

→ calculations DRs

→ component choices

→ schematic

→ exp. setup

→ results

• VARIABLE R

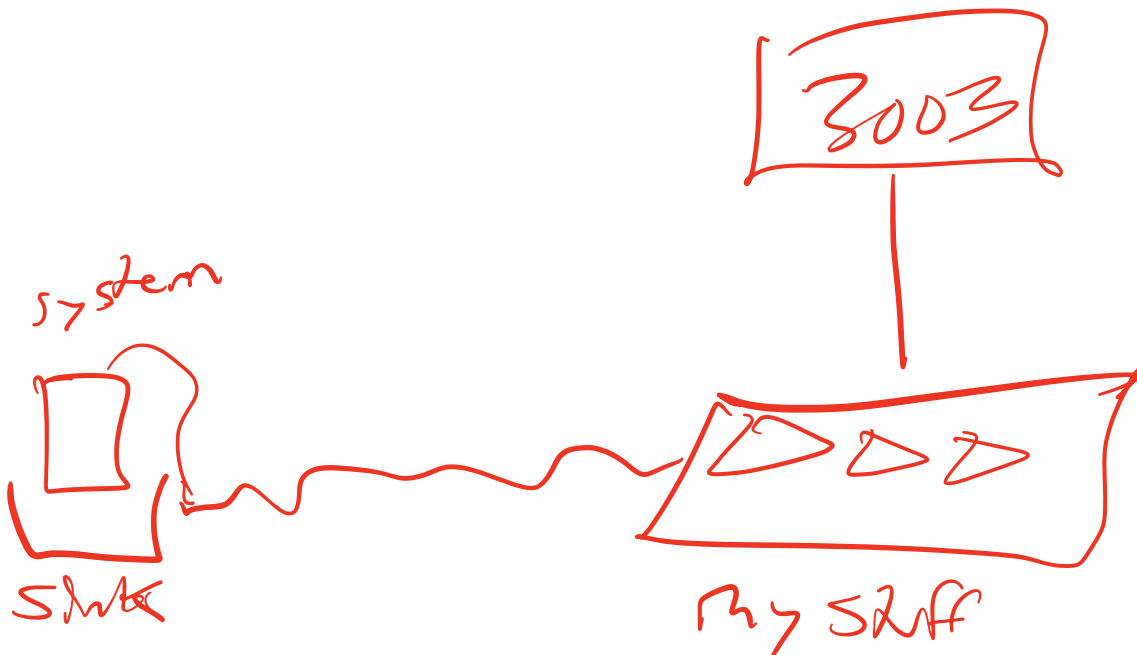
• "PV Panel"

• Step responses

→ draws / em.

• 22 AWG → 5A ✓

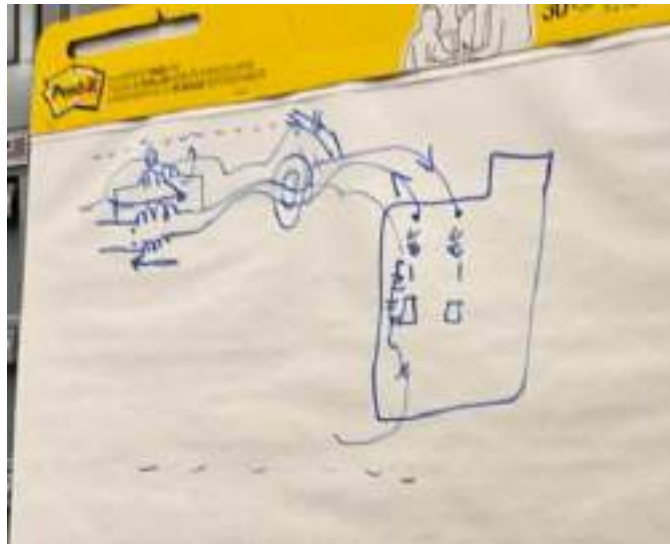
• PCB trace 1.27 mm → 0.050" ✓
for 3A

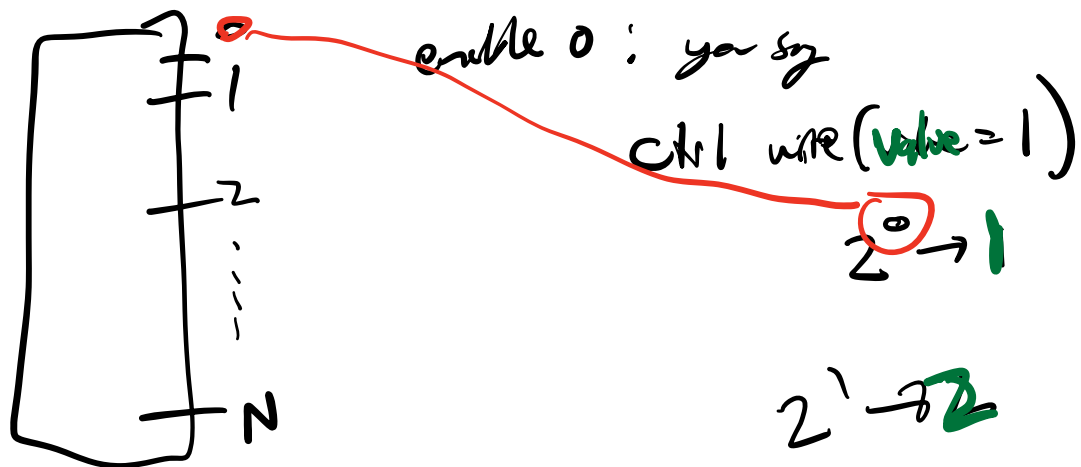


• 2N5 — Common Mode
Choke

• 2N8
app
nd
bottom
Share

oo
oe





addr / value

$$2^1 \rightarrow 2$$

$$2^2 \rightarrow 4$$

$$2^3 \rightarrow 8$$

⋮

if —
ctrl reg write ()

if

